

# LF198JAN

## Monolithic Sample-and-Hold Circuits

### General Description

The LF198 is a monolithic sample-and-hold circuit which utilizes BI-FET technology to obtain ultra-high dc accuracy with fast acquisition of signal and low droop rate. Operating as a unity gain follower, dc gain accuracy is 0.002% typical and acquisition time is as low as 6  $\mu$ s to 0.01%. A bipolar input stage is used to achieve low offset voltage and wide bandwidth. Input offset adjust is accomplished with a single pin, and does not degrade input offset drift. The wide bandwidth allows the LF198 to be included inside the feedback loop of 1 MHz op amps without having stability problems. Input impedance of  $10^{10}\Omega$  allows high source impedances to be used without degrading accuracy.

P-channel junction FET's are combined with bipolar devices in the output amplifier to give droop rates as low as 5 mV/min with a 1  $\mu$ F hold capacitor. The JFET's have much lower noise than MOS devices used in previous designs and do not exhibit high temperature instabilities. The overall design guarantees no feed-through from input to output in the hold mode, even for input signals equal to the supply voltages.

### Features

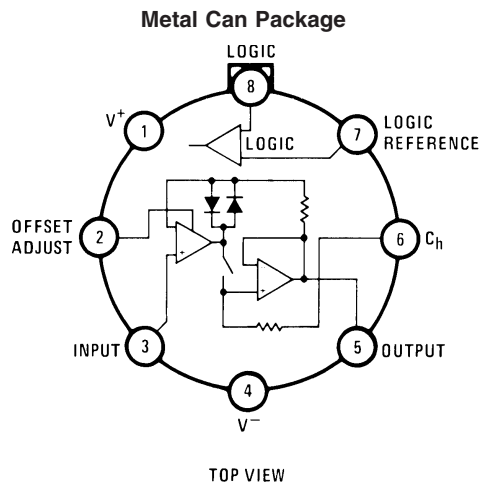
- Operates from  $\pm 5$ V to  $\pm 18$ V supplies
- Less than 10  $\mu$ s acquisition time
- TTL, PMOS, CMOS compatible logic input
- 0.5 mV typical hold step at  $C_h = 0.01 \mu$ F
- Low input offset
- 0.002% gain accuracy
- Low output noise in hold mode
- Input characteristics do not change during hold mode
- High supply rejection ratio in sample or hold
- Wide bandwidth
- Space Qualified

Logic inputs on the LF198 are fully differential with low input current, allowing direct connection to TTL, PMOS, and CMOS. Differential threshold is 1.4V. The LF198 will operate from  $\pm 5$ V to  $\pm 18$ V supplies.

### Ordering Information

NSC Part Number	JAN Part Number	NSC Package Number	Package Description
JL198BGA	JM38510/12501BGA	H08C	8LD Metal Can
JL198SGA	JM38510/12501SGA	H08C	8LD Metal Can

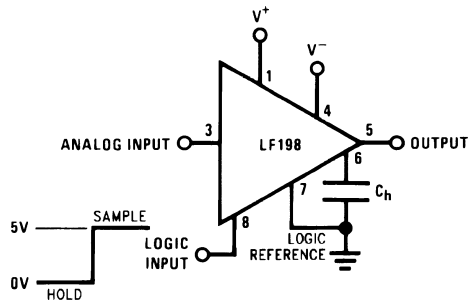
### Connection Diagrams



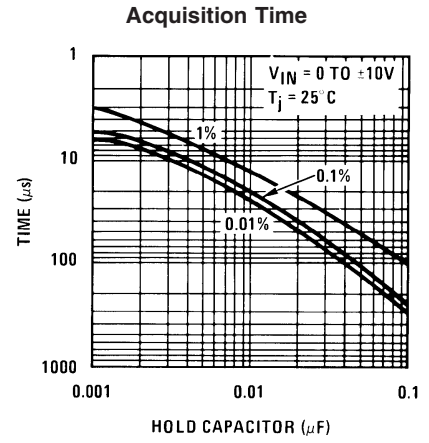
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See NS Package Number H08C

# Typical Connection and Performance Curve

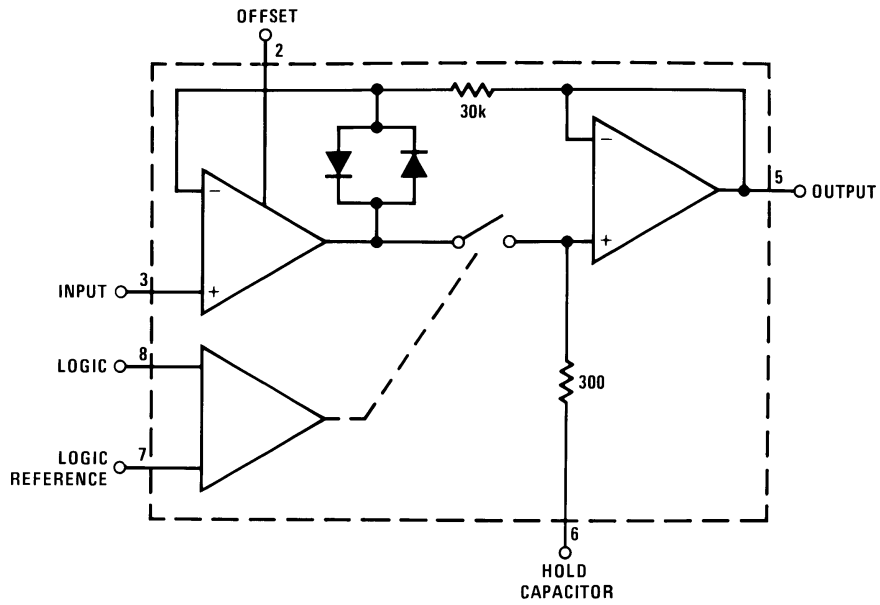


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## Functional Diagram



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## Absolute Maximum Ratings (Note 1)

Supply Voltage	±18V
Power Dissipation (Package Limitation) (Note 2)	500 mW
Operating Ambient Temperature Range	$-55^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$
Storage Temperature Range	$-65^{\circ}\text{C}$ to $+150^{\circ}\text{C}$
Maximum Junction Temperature ( $T_{Jmax}$ )	$+150^{\circ}\text{C}$
Input Voltage	Equal to Supply Voltage
Logic To Logic Reference Differential Voltage (Note 3)	+7V, -30V
Output Short Circuit Duration	Indefinite
Hold Capacitor Short Circuit Duration	10 sec
Lead Temperature (Soldering, 10 sec.)	$300^{\circ}\text{C}$
Thermal Resistance	
$\theta_{JA}$	
Metal Can (Still Air @ 0.5W)	$160^{\circ}\text{C}/\text{W}$
Metal Can (500 LF/Min Air Flow @ 0.5W)	$84^{\circ}\text{C}/\text{W}$
$\theta_{JC}$	
Metal Can	$48^{\circ}\text{C}/\text{W}$
ESD Tolerance (Note 7)	500V

## Quality Conformance Inspection

Mil-Std-883, Method 5005 — Group A

Subgroup	Description	Temperature ( $^{\circ}\text{C}$ )
1	Static tests at	$+25^{\circ}\text{C}$
2	Static tests at	$+125^{\circ}\text{C}$
3	Static tests at	$-55^{\circ}\text{C}$
4	Dynamic tests at	$+25^{\circ}\text{C}$
5	Dynamic tests at	$+125^{\circ}\text{C}$
6	Dynamic tests at	$-55^{\circ}\text{C}$
7	Functional tests at	$+25^{\circ}\text{C}$
8A	Functional tests at	$+125^{\circ}\text{C}$
8B	Functional tests at	$-55^{\circ}\text{C}$
9	Switching tests at	$+25^{\circ}\text{C}$
10	Switching tests at	$+125^{\circ}\text{C}$
11	Switching tests at	$-55^{\circ}\text{C}$

## Electrical Characteristics

### DC Parameters

Symbol	Parameter	Conditions	Notes	Min	Max	Unit	Sub-groups		
$V_{IO}$	Input Offset Voltage	$+V_{CC} = 3.5V, -V_{CC} = -26.5V,$ $V_{CM} = 11.5V$		-3.0	3.0	mV	1		
				-5.0	5.0	mV	2, 3		
		$+V_{CC} = 26.5V, -V_{CC} = -3.5V,$ $V_{CM} = -11.5V$		-3.0	3.0	mV	1		
				-5.0	5.0	mV	2, 3		
		$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$		-3.0	3.0	mV	1		
				-5.0	5.0	mV	2, 3		
$I_{IB}$	Input Bias Current	$+V_{CC} = 3.5V, -V_{CC} = -26.5V,$ $V_{CM} = 11.5V$		-1.0	25	nA	1		
				-25	75	nA	2, 3		
		$+V_{CC} = 26.5V, -V_{CC} = -3.5V, V_{CM}$ $= -11.5V$		-1.0	25	nA	1		
				-25	75	nA	2, 3		
		$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$		-1	25	nA	1		
				-25	75	nA	2, 3		
$Z_i$	Input Impedance	$+V_{CC} = 3.5V$ to $26.6V,$ $-V_{CC} = -26.5V$ to $-3.5V,$ $V_{CM} = 11.5V$ to $-11.5V$		2.0		G $\Omega$	1		
				1.0		G $\Omega$	2, 3		
		$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$		6.0		mV	1, 2, 3		
					-6.0	mV	1, 2, 3		
		PSRR+	Power Supply Rejection Ratio	$-V_{CC} = -18V,$ $+V_{CC} = 18V$ to $12V$		80		dB	1, 2, 3
						80		dB	1, 2, 3
$I_{CC}$	Supply Current	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$		1.0	5.5	mA	1,2		
				1.0	6.5	mA	3		
$A_E$	Gain Error	$+V_{CC} = 3.5V$ to $26.5V,$ $-V_{CC} = -26.5V$ to $-3.5V,$ $V_{CM} = -11.5V$ to $11.5V$		-0.005	0.005	%	1		
				-0.02	0.02	%	2, 3		
		$+V_{CC} = 3V$ to $7V,$ $-V_{CC} = -7V$ to $-3V,$ $V_{CM} = -2V$ to $2V$		-0.02	0.02	%	1		
				-0.04	0.04	%	2, 3		
$R_{SC}$	Series Charge Resistance	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$		75	400	$\Omega$	1, 2, 3		
$I_{IH}$ (a)	Logical 1 Input Current	$+V_{CC} = 8.5V, -V_{CC} = -21.5V$		0	10	$\mu A$	1, 2, 3		
$I_{IH}$ (b)	Logical 1 Input Current	$+V_{CC} = 8.5V, -V_{CC} = -21.5V$		0	10	$\mu A$	1, 2, 3		
$I_{IL}$ (a)	Logical 0 Input Current	$+V_{CC} = 21.5V, -V_{CC} = -8.5V$		-1.0	1.0	$\mu A$	1, 2, 3		
$I_{IL}$ (b)	Logical 0 Input Current	$+V_{CC} = 21.5V, -V_{CC} = -8.5V$		-1.0	1.0	$\mu A$	1, 2, 3		
$I_{OS+}$	Output Short Circuit Current	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$		-25		mA	1, 2, 3		

**Electrical Characteristics** (Continued)**DC Parameters** (Continued)

Symbol	Parameter	Conditions	Notes	Min	Max	Unit	Sub-groups
$I_{OS^-}$	Output Short Circuit Current	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$			25	mA	1, 2, 3
$I_{CH^+}$	Hold Capacitor Charge Current	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$			-3.0	mA	1
					-2.0	mA	2, 3
$I_{CH^-}$	Hold Capacitor Charge Current	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$		3.0		mA	1
				2.0		mA	2, 3
$V_{TH(H)}$	Differential Logic Threshold	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$ Logic = 2.0V, Logic Ref = 2.0V		1.0		mA	1, 2, 3
$V_{TH(L)}$	Differential Logic Threshold	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$ Logic = 0.8V, Logic Ref = 2.0V		-10	10	$\mu A$	1, 2, 3
$I_{HL^+}$	Hold Mode Leakage Current	$+V_{CC} = 3.5V, -V_{CC} = -26.5V,$ $V_{CM} = -11.5V$	(Note 5)	-0.100	0.100	nA	1
				-50	50	nA	2
$I_{HL^-}$	Hold Mode Leakage Current	$+V_{CC} = 26.5V, -V_{CC} = -3.5V,$ $V_{CM} = 11.5V$	(Note 5)	-0.100	0.100	nA	1
				-50	50	nA	2
$Z_O$	Output Impedance	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V$			2.0	$\Omega$	1, 2, 3
$V_{HS}$	(HOLD) Step Voltage	$+V_{CC} = 3.5V, -V_{CC} = -26.5V,$ $V_{CM} = 11.5V$	(Note 4)	-2.0	2.0	mV	1
				-5.0	5.0	mV	2, 3
		$+V_{CC} = 26.5V, -V_{CC} = -3.5V,$ $V_{CM} = -11.5V$	(Note 4)	-2.0	2.0	mV	1
				-5.0	5.0	mV	2, 3
$F_{RR}$	Feedthrough Rejection Ratio	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V, V_I = 0V \text{ to } 11.5V$		86		dB	1
				80		dB	2, 3
		$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V, V_I = 11.5V \text{ to } 0V$		86		dB	1
				80		dB	2, 3
		$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V, V_I = 0V \text{ to } -11.5V$		86		dB	1
				80		dB	2, 3
$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_{CM} = 0V, V_I = -11.5V \text{ to } 0V$		86		dB	1		
		80		dB	2, 3		

**AC/DC Parameters**

Symbol	Parameter	Conditions	Notes	Min	Max	Unit	Sub-groups
$\Delta V_{IO} / \Delta T$	Input Offset Voltage Temp Sensitivity			-20	20	$\mu V/^\circ C$	8A, 8B
$T_{AQ}$	Aquisition Time	$+V_{CC} = 15V, -V_{CC} = -15V$			25	$\mu S$	7
$T_{AP}$	Aperture Time	$+V_{CC} = 15V, -V_{CC} = -15V$			300	nS	7
$T_S$	Settling Time	$+V_{CC} = 15V, -V_{CC} = -15V$			1.5	$\mu S$	7
$F_{RR AC}$	Feedthrough Rejection Ratio	$+V_{CC} = 15V, -V_{CC} = -15V,$ $V_I = 20V_{pp}$		86		dB	7
$TR_{TS}$	Transient Response (settling time)	$+V_{CC} = 3.5V, -V_{CC} = -26.5V,$ $V_I = 100mV \text{ pulse}$			2.5	$\mu S$	7
		$+V_{CC} = 26.5V, -V_{CC} = -3.5V,$ $V_I = 100mV \text{ pulse}$			2.5	$\mu S$	7

## AC/DC Parameters (Continued)

Symbol	Parameter	Conditions	Notes	Min	Max	Unit	Sub-groups
TR <sub>OS</sub>	Transient Response (overshoot)	+V <sub>CC</sub> = 3.5V, -V <sub>CC</sub> = -26.5V, V <sub>I</sub> = 100mV pulse			40	%	7
		+V <sub>CC</sub> = 26.5V, -V <sub>CC</sub> = -3.5V, V <sub>I</sub> = 100mV pulse			40	%	7
en <sub>H</sub>	Noise	+V <sub>CC</sub> = 15V, -V <sub>CC</sub> = -15V			10	mV <sub>RMS</sub>	7
en <sub>S</sub>	Noise	+V <sub>CC</sub> = 15V, -V <sub>CC</sub> = -15V			10	mV <sub>RMS</sub>	7

## DC Parameters: Drift Values

Delta calculations performed on S-Level devices at group B, subgroup 5 ONLY.

Symbol	Parameters	Conditions	Notes	Min	Max	Unit	Sub-groups
V <sub>IO</sub>	Input Offset Voltage	+V <sub>CC</sub> = 15V, -V <sub>CC</sub> = -15V, V <sub>CM</sub> = 0V		-0.5	0.5	mV	1
I <sub>IB</sub>	Input Bias Current	+V <sub>CC</sub> = 15V, -V <sub>CC</sub> = -15V, V <sub>CM</sub> = 0V		-2.5	2.5	nA	1

**Note 1:** "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics. The guaranteed specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions

**Note 2:** The maximum power dissipation must be derated at elevated temperatures and is dictated by T<sub>JMAX</sub>, θ<sub>JA</sub>, and the ambient temperature, T<sub>A</sub>. The maximum allowable power dissipation at any temperature is P<sub>D</sub> = (T<sub>JMAX</sub> - T<sub>A</sub>)/θ<sub>JA</sub>, or the number given in the Absolute Maximum Ratings, whichever is lower. .

**Note 3:** Although the differential voltage may not exceed the limits given, the common-mode voltage on the logic pins may be equal to the supply voltages without causing damage to the circuit. For proper logic operation, however, one of the logic pins must always be at least 2V below the positive supply and 3V above the negative supply.

**Note 4:** Hold step is sensitive to stray capacitive coupling between input logic signals and the hold capacitor. 1 pF, for instance, will create an additional 0.5 mV step with a 5V logic swing and a 0.01µF hold capacitor. Magnitude of the hold step is inversely proportional to hold capacitor value.

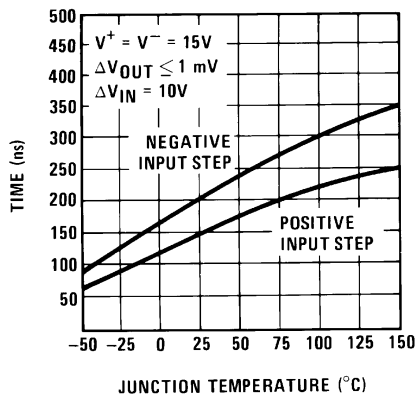
**Note 5:** Leakage current is measured at a junction temperature of 25°C. The effects of junction temperature rise due to power dissipation or elevated ambient can be calculated by doubling the 25°C value for each 11°C increase in chip temperature. Leakage is guaranteed over full input signal range.

**Note 6:** See Definition of Terms

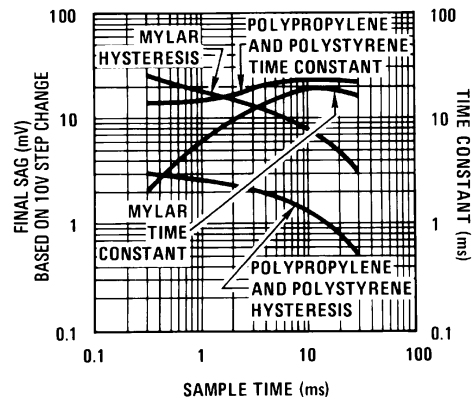
**Note 7:** Human body model, 100pF discharged through 1.5KΩ

## Typical Performance Characteristics

Aperture Time  
(Note 6)

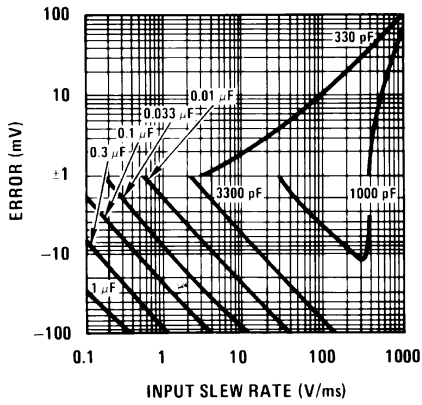


Dielectric Absorption  
Error in Hold Capacitor



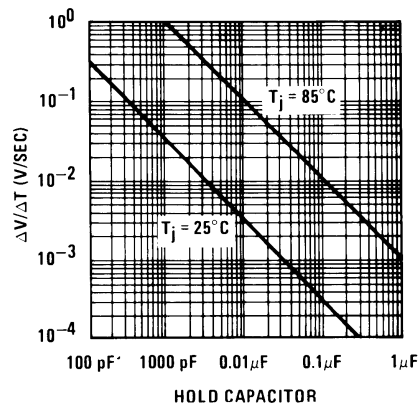
Typical Performance Characteristics (Continued)

Dynamic Sampling Error  
(Note 6)



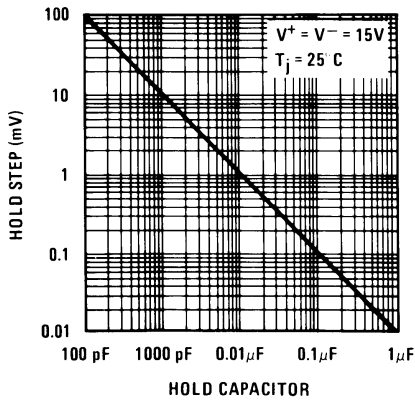
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Output Droop Rate



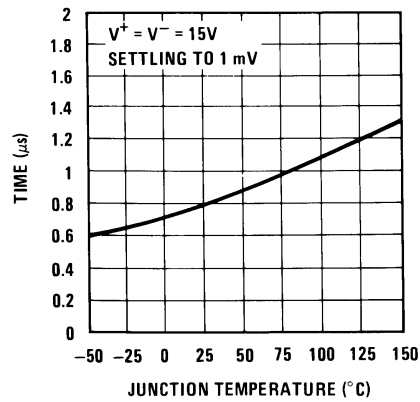
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Hold Step  
(Note 6)



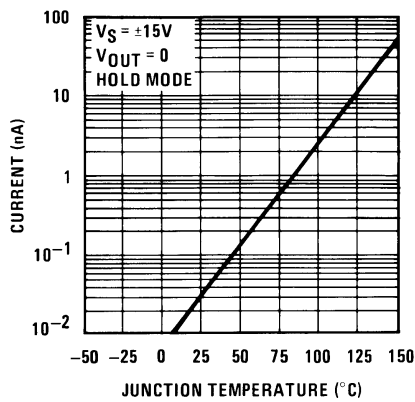
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"Hold" Settling Time  
(Note 6)



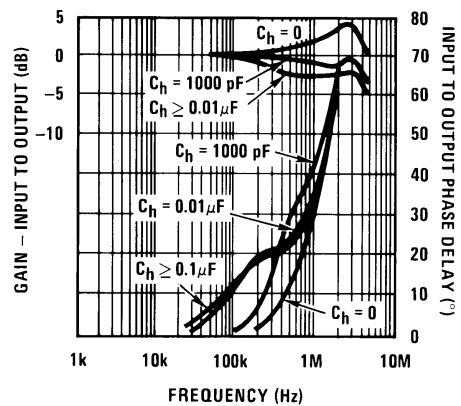
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Leakage Current into Hold Capacitor



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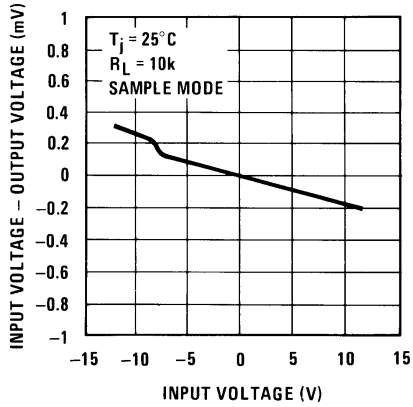
Phase and Gain (Input to Output, Small Signal)



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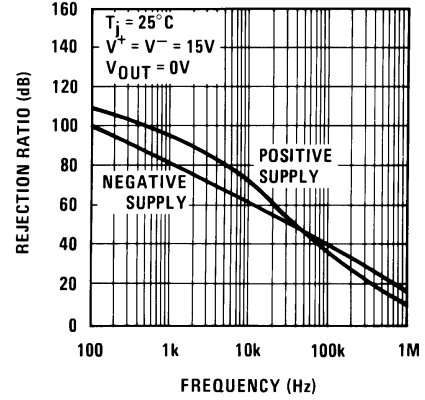
# Typical Performance Characteristics (Continued)

**Gain Error**



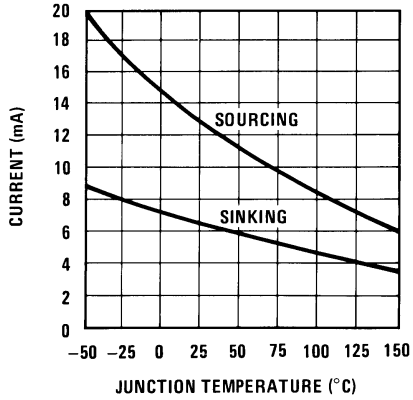
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**Power Supply Rejection**



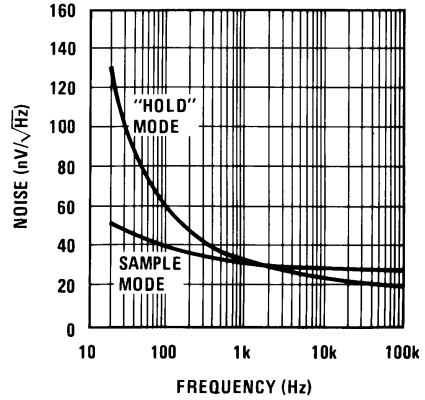
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**Output Short Circuit Current**



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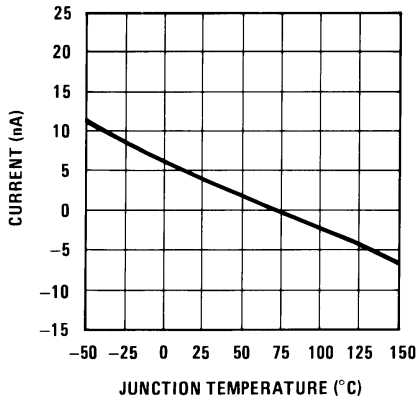
**Output Noise**



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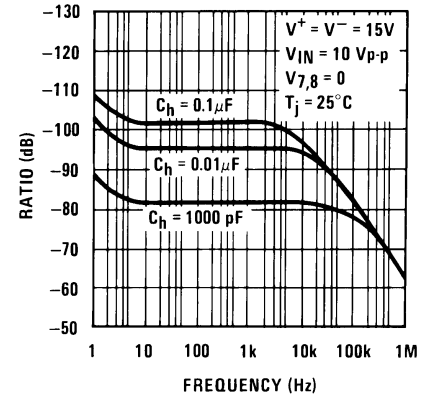
Note 8: See Definition

**Input Bias Current**



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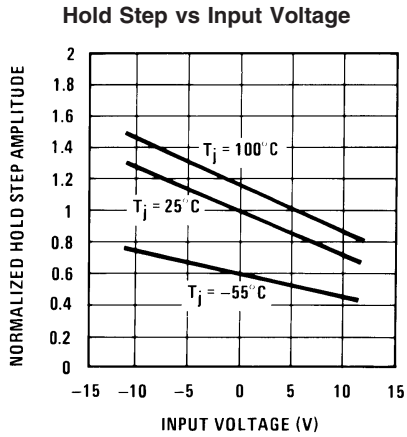
**Feedthrough Rejection Ratio (Hold Mode)**



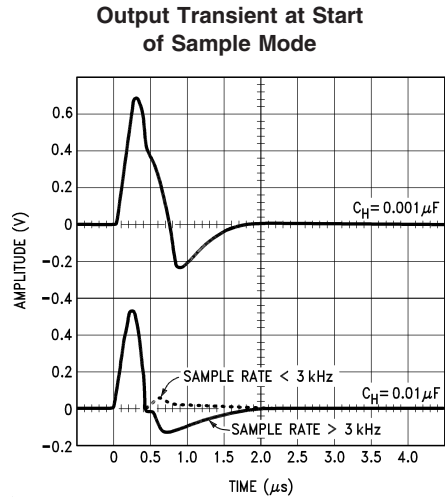
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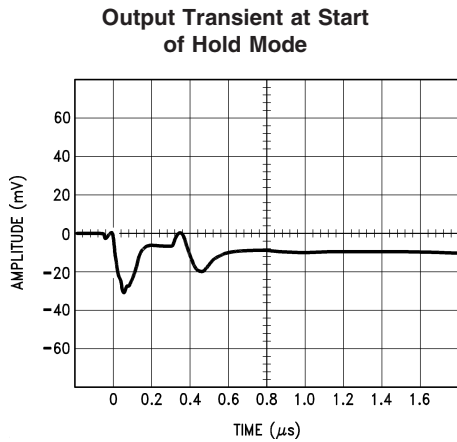
# Typical Performance Characteristics (Continued)



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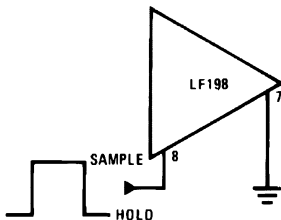
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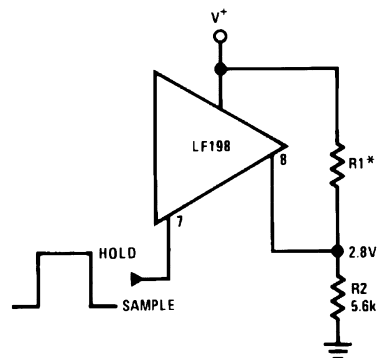
## Logic Input Configurations

TTL & CMOS  
 $3\text{V} \leq V_{\text{LOGIC}} (\text{Hi State}) \leq 7\text{V}$



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Threshold = 1.4V



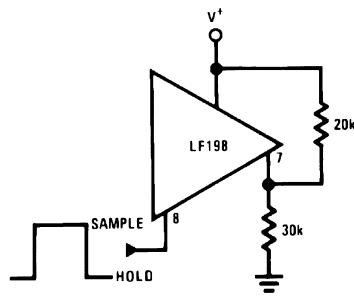
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Threshold =  $1.4\text{V} \cdot \text{Select for } 2.8\text{V}$  at pin 8

## Logic Input Configurations (Continued)

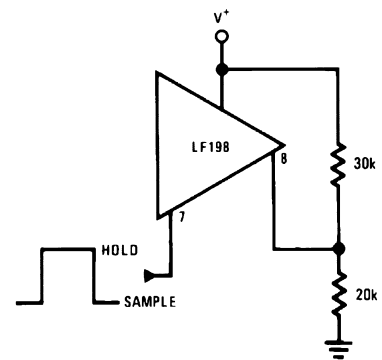
### CMOS

$$7V \leq V_{\text{LOGIC}} (\text{Hi State}) \leq 15V$$



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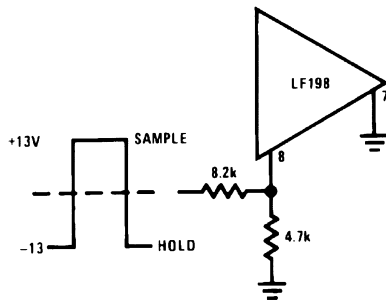
$$\text{Threshold} = 0.6 (V^+) + 1.4V$$



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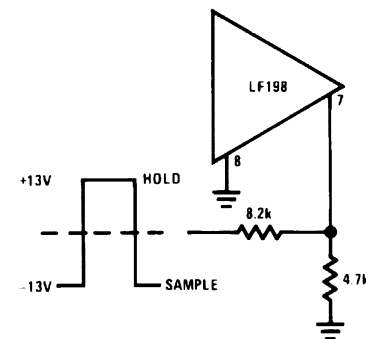
$$\text{Threshold} = 0.6 (V^+) - 1.4V$$

### Op Amp Drive



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$$\text{Threshold} \approx +4V$$



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$$\text{Threshold} = -4V$$

## Application Hints

### HOLD CAPACITOR

Hold step, acquisition time, and droop rate are the major trade-offs in the selection of a hold capacitor value. Size and cost may also become important for larger values. Use of the curves included with this data sheet should be helpful in selecting a reasonable value of capacitance. Keep in mind that for fast repetition rates or tracking fast signals, the capacitor drive currents may cause a significant temperature rise in the LF198.

A significant source of error in an accurate sample and hold circuit is dielectric absorption in the hold capacitor. A mylar cap, for instance, may "sag back" up to 0.2% after a quick change in voltage. A long sample time is required before the circuit can be put back into the hold mode with this type of capacitor. Dielectrics with very low hysteresis are polystyrene, polypropylene, and Teflon. Other types such as mica and polycarbonate are not nearly as good. The advantage of polypropylene over polystyrene is that it extends the maximum ambient temperature from 85°C to 100°C. Most ceramic capacitors are unusable with > 1% hysteresis. Ceramic "NPO" or "COG" capacitors are now available for 125°C operation and also have low dielectric absorption. For more exact data, see the curve *Dielectric Absorption Error*.

The hysteresis numbers on the curve are final values, taken after full relaxation. The hysteresis error can be significantly reduced if the output of the LF198 is digitized quickly after the hold mode is initiated. The hysteresis relaxation time constant in polypropylene, for instance, is 10—50 ms. If A-to-D conversion can be made within 1 ms, hysteresis error will be reduced by a factor of ten.

### DC AND AC ZEROING

DC zeroing is accomplished by connecting the offset adjust pin to the wiper of a 1 kΩ potentiometer which has one end tied to V<sup>+</sup> and the other end tied through a resistor to ground. The resistor should be selected to give ≈0.6 mA through the 1k potentiometer.

AC zeroing (hold step zeroing) can be obtained by adding an inverter with the adjustment pot tied input to output. A 10 pF capacitor from the wiper to the hold capacitor will give ±4 mV hold step adjustment with a 0.01 μF hold capacitor and 5V logic supply. For larger logic swings, a smaller capacitor (< 10 pF) may be used.

### LOGIC RISE TIME

For proper operation, logic signals into the LF198 must have a minimum dV/dt of 1.0 V/μs. Slower signals will cause excessive hold step. If a R/C network is used in front of the

## Application Hints (Continued)

logic input for signal delay, calculate the slope of the waveform at the threshold point to ensure that it is at least  $1.0 \text{ V}/\mu\text{s}$ .

### SAMPLING DYNAMIC SIGNALS

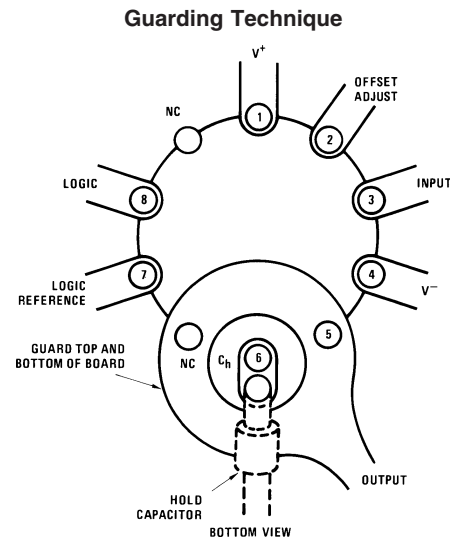
Sample error to moving input signals probably causes more confusion among sample-and-hold users than any other parameter. The primary reason for this is that many users make the assumption that the sample and hold amplifier is truly locked on to the input signal while in the sample mode. In actuality, there are finite phase delays through the circuit creating an input-output differential for fast moving signals. In addition, although the output may have settled, the hold capacitor has an additional lag due to the  $300\Omega$  series resistor on the chip. This means that at the moment the "hold" command arrives, the hold capacitor voltage may be somewhat different than the actual analog input. The effect of these delays is opposite to the effect created by delays in the logic which switches the circuit from sample to hold. For example, consider an analog input of  $20 \text{ Vp-p}$  at  $10 \text{ kHz}$ . Maximum  $dV/dt$  is  $0.6 \text{ V}/\mu\text{s}$ . With no analog phase delay and  $100 \text{ ns}$  logic delay, one could expect up to  $(0.1 \mu\text{s})(0.6 \text{ V}/\mu\text{s}) = 60 \text{ mV}$  error if the "hold" signal arrived near maximum  $dV/dt$  of the input. A positive-going input would give a  $+60 \text{ mV}$  error. Now assume a  $1 \text{ MHz}$  ( $3 \text{ dB}$ ) bandwidth for the overall analog loop. This generates a phase delay of  $160 \text{ ns}$ . If the hold capacitor sees this exact delay, then error due to analog delay will be  $(0.16 \mu\text{s})(0.6 \text{ V}/\mu\text{s}) = -96 \text{ mV}$ . Total output error is  $+60 \text{ mV}$  (digital)  $-96 \text{ mV}$  (analog) for a total of  $-36 \text{ mV}$ . To add to the confusion, analog delay is proportioned to hold capacitor value while digital delay remains constant. A family of curves (dynamic sampling error) is included to help estimate errors.

A curve labeled *Aperture Time* has been included for sampling conditions where the input is steady during the sampling period, but may experience a sudden change nearly coincident with the "hold" command. This curve is based on a  $1 \text{ mV}$  error fed into the output.

A second curve, *Hold Settling Time* indicates the time required for the output to settle to  $1 \text{ mV}$  after the "hold" command.

### DIGITAL FEEDTHROUGH

Fast rise time logic signals can cause hold errors by feeding externally into the analog input at the same time the amplifier is put into the hold mode. To minimize this problem, board layout should keep logic lines as far as possible from the analog input and the  $C_h$  pin. Grounded guarding traces may also be used around the input line, especially if it is driven from a high impedance source. Reducing high amplitude logic signals to  $2.5\text{V}$  will also help.

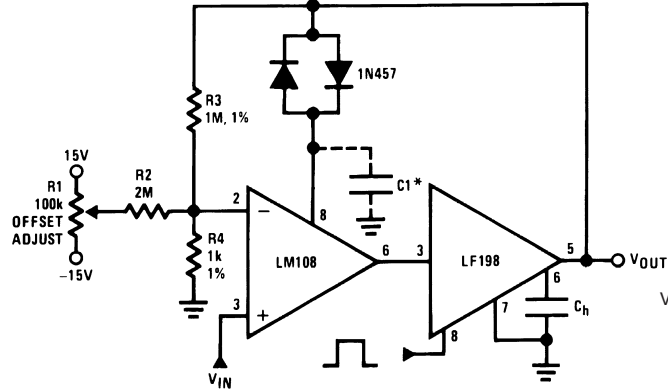


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**Use 10-pin layout. Guard around  $C_h$  is tied to output.**

# Typical Applications

X1000 Sample & Hold

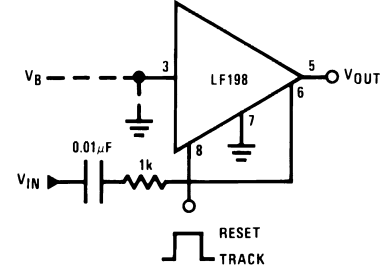


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\*For lower gains, the LM108 must be frequency compensated

$$\text{Use } \approx \frac{100}{A_v} \text{ pF from comp 2 to ground}$$

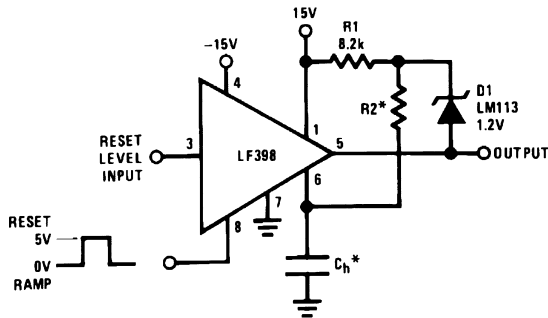
Sample and Difference Circuit  
(Output Follows Input in Hold Mode)



20128140

$$V_{OUT} = V_B + \Delta V_{IN}(\text{HOLD MODE})$$

Ramp Generator with Variable Reset Level

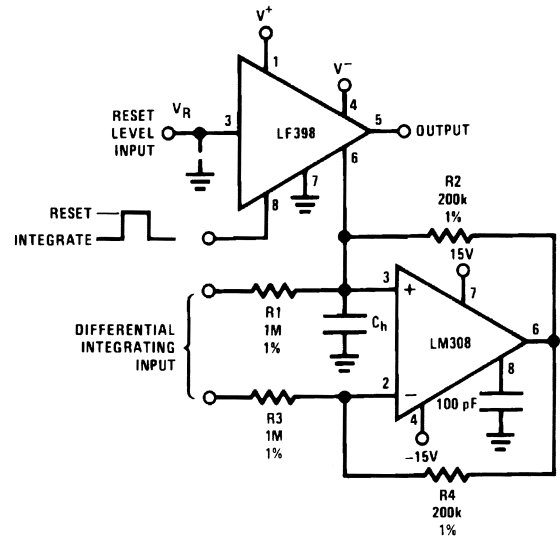


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$$\text{*Select for ramp rate } \frac{\Delta V}{\Delta T} = \frac{1.2V}{(R2)(C_h)}$$

R2 ≥ 10k

Integrator with Programmable Reset Level

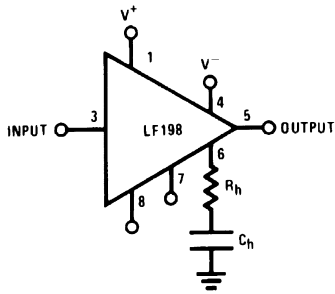


20128143

$$V_{OUT}(\text{Hold Mode}) = \left[ \frac{1}{(R1)(C_h)} \int_0^t V_{IN} dt \right] + [V_R]$$

# Typical Applications (Continued)

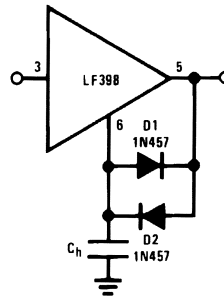
Output Holds at Average of Sampled Input



20128146

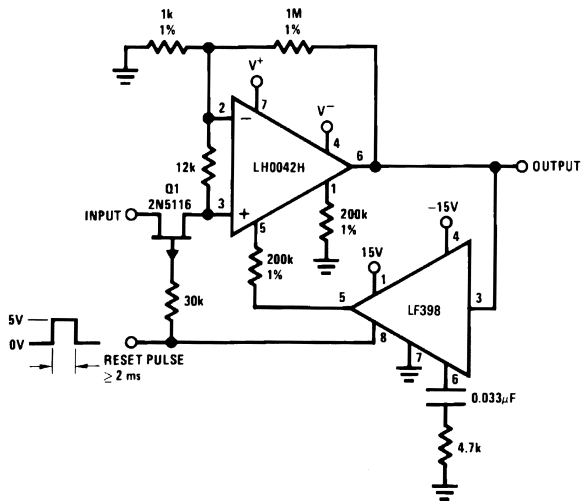
$$\text{Select } (R_h) (C_h) \geq \frac{1}{2\pi f_{IN} (\text{Min})}$$

Increased Slew Current



20128147

Reset Stabilized Amplifier (Gain of 1000)



20128149

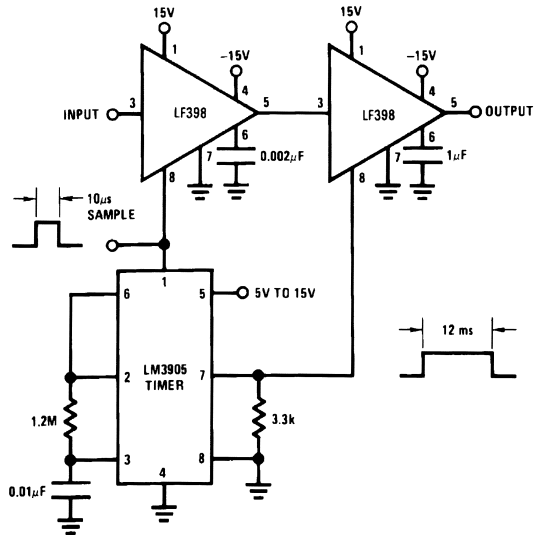
$$V_{OS} \leq 20\mu\text{V (No trim)}$$

$$Z_{IN} \approx 1\text{ M}\Omega$$

$$\frac{\Delta V_{OS}}{\Delta t} \approx 30\mu\text{V/sec}$$

$$\frac{\Delta V_{OS}}{\Delta T} \approx 0.1\mu\text{V}/^\circ\text{C}$$

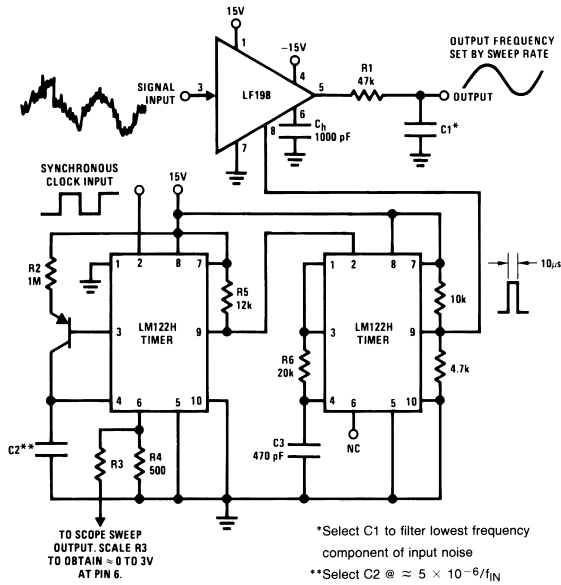
Fast Acquisition, Low Droop Sample & Hold



20128150

# Typical Applications (Continued)

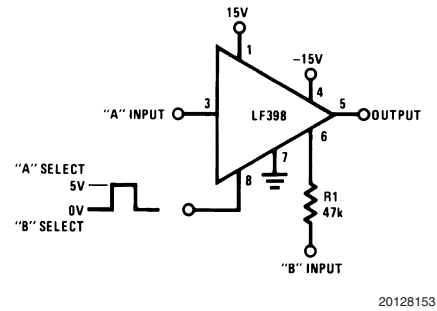
## Synchronous Correlator for Recovering Signals Below Noise Level



20128152

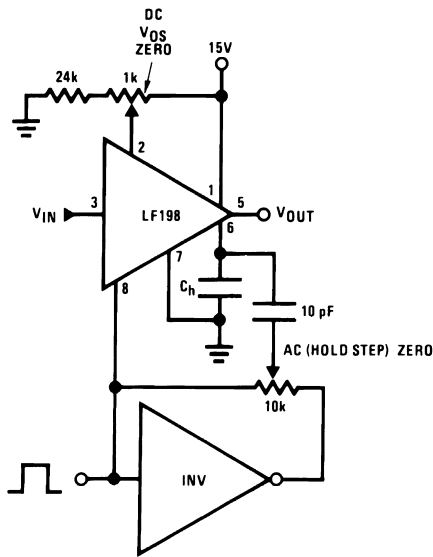
	A	B
Gain	$1 \pm 0.02\%$	$1 \pm 0.2\%$
$Z_{IN}$	$10^{10}\Omega$	47 k $\Omega$
BW	$\approx 1$ MHz	$\approx 400$ kHz
Crosstalk @ 1 kHz	-90 dB	-90 dB
Offset	$\leq 6$ mV	$\leq 75$ mV

## 2-Channel Switch



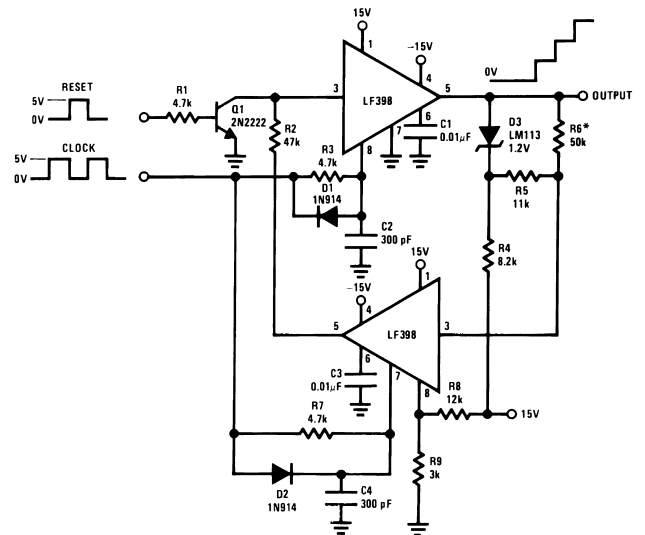
20128153

## DC & AC Zeroing



20128159

## Staircase Generator

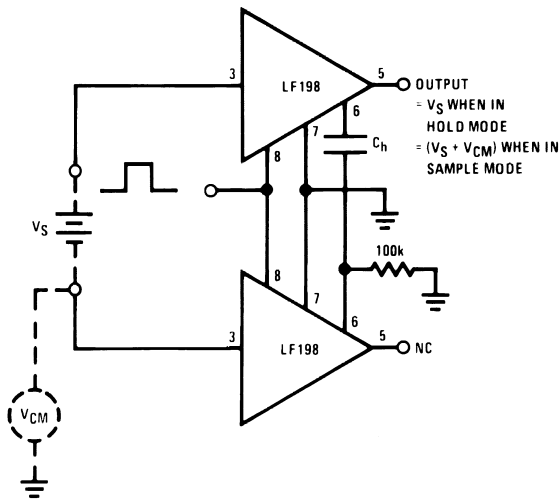


\*Select for step height  
50k  $\rightarrow$   $\approx 1$ V Step

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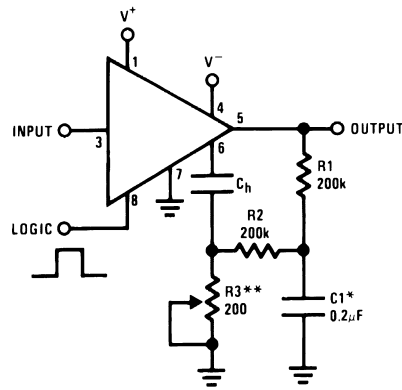
## Typical Applications (Continued)

### Differential Hold



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### Capacitor Hysteresis Compensation



20128156

\*Select for time constant  $C1 = \frac{\tau}{100k}$

\*\*Adjust for amplitude

## Definition of Terms

**Hold Step:** The voltage step at the output of the sample and hold when switching from sample mode to hold mode with a steady (dc) analog input voltage. Logic swing is 5V.

**Acquisition Time:** The time required to acquire a new analog input voltage with an output step of 10V. Note that acquisition time is not just the time required for the output to settle, but also includes the time required for all internal nodes to settle so that the output assumes the proper value when switched to the hold mode.

**Gain Error:** The ratio of output voltage swing to input voltage swing in the sample mode expressed as a per cent difference.

**Hold Settling Time:** The time required for the output to settle within 1 mV of final value after the "hold" logic command.

**Dynamic Sampling Error:** The error introduced into the held output due to a changing analog input at the time the hold command is given. Error is expressed in mV with a given hold capacitor value and input slew rate. Note that this error term occurs even for long sample times.

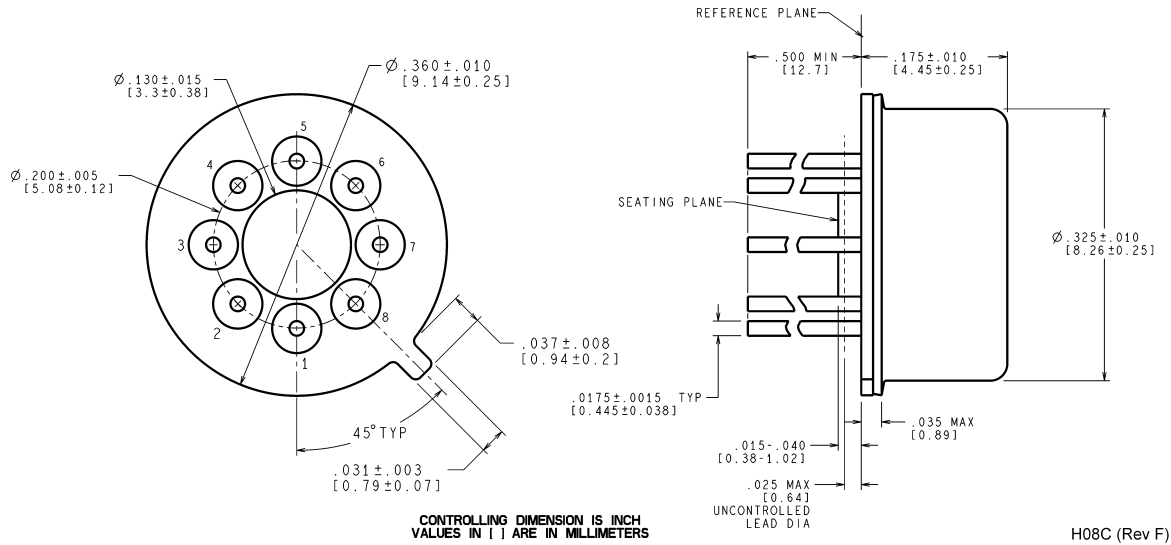
**Aperture Time:** The delay required between "Hold" command and an input analog transition, so that the transition does not affect the held output.

**Revision History Section**

Date Released	Revision	Section	Originator	Changes
02/25/05	A	New release, Corporate format	L. Lytle	1 MDS converted to corp. datasheet format. MJLF198-X Rev 2B0 MDS to be archived.



**Physical Dimensions** inches (millimeters) unless otherwise noted



CONTROLLING DIMENSION IS INCH  
VALUES IN [ ] ARE IN MILLIMETERS

H08C (Rev F)

**Metal Can Package (H)  
NS Package Number H08C**

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