

# OP27A, OP27C

## LOW-NOISE HIGH-SPEED PRECISION OPERATIONAL AMPLIFIERS

SLOS100D – FEBRUARY 1989 – REVISED JANUARY 2007

- Replacements for ADI, PMI and LTC OP27 Series

### Features of OP27A and OP27C:

- **Maximum Equivalent Input Noise Voltage:**  
 3.8 nV/√Hz at 1 kHz  
 5.5 nV/√Hz at 10 kHz
- **Very Low Peak-to-Peak Noise Voltage at 0.1 Hz to 10 Hz . . . 80 nV Typ**
- **Low Input Offset Voltage**  
 OP27A . . . 25 μV Max  
 OP27C . . . 100 μV Max
- **High Voltage Amplification**  
 OP27A . . . 1 V/μV Min  
 OP27C . . . 0.7 V/μV Min

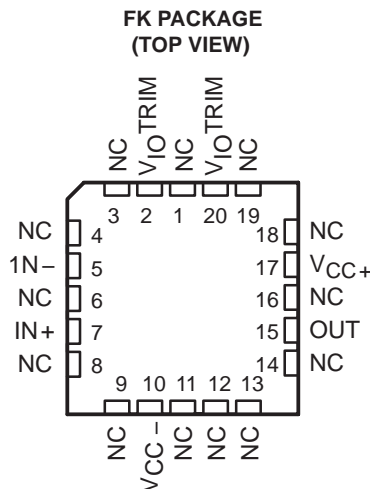
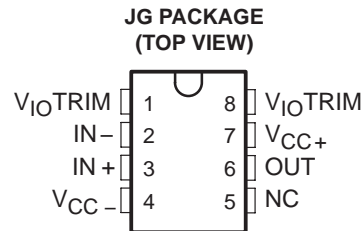
### description

The OP27 operational amplifiers combine outstanding noise performance with excellent precision and high-speed specifications. The wideband noise is only 3 nV/√Hz and with the 1/f noise corner at 2.7 Hz, low noise is maintained for all low-frequency applications.

The outstanding characteristics of the OP27 make these devices excellent choices for low-noise amplifier applications requiring precision performance and reliability.

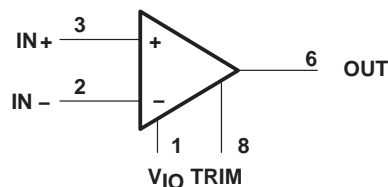
The OP27 series is compensated for unity gain.

The OP27A and OP27C are characterized for operation over the full military temperature range of -55°C to 125°C.



NC – No internal connection

### symbol



Pin numbers are for the JG packages.

### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	STABLE GAIN	PACKAGE	
			CERAMIC DIP (JG)	CHIP CARRIER (FK)
-55°C to 125°C	25 μV	1	OP27AJG	OP27AFK
	100 μV	1	OP27CJG	—



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PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

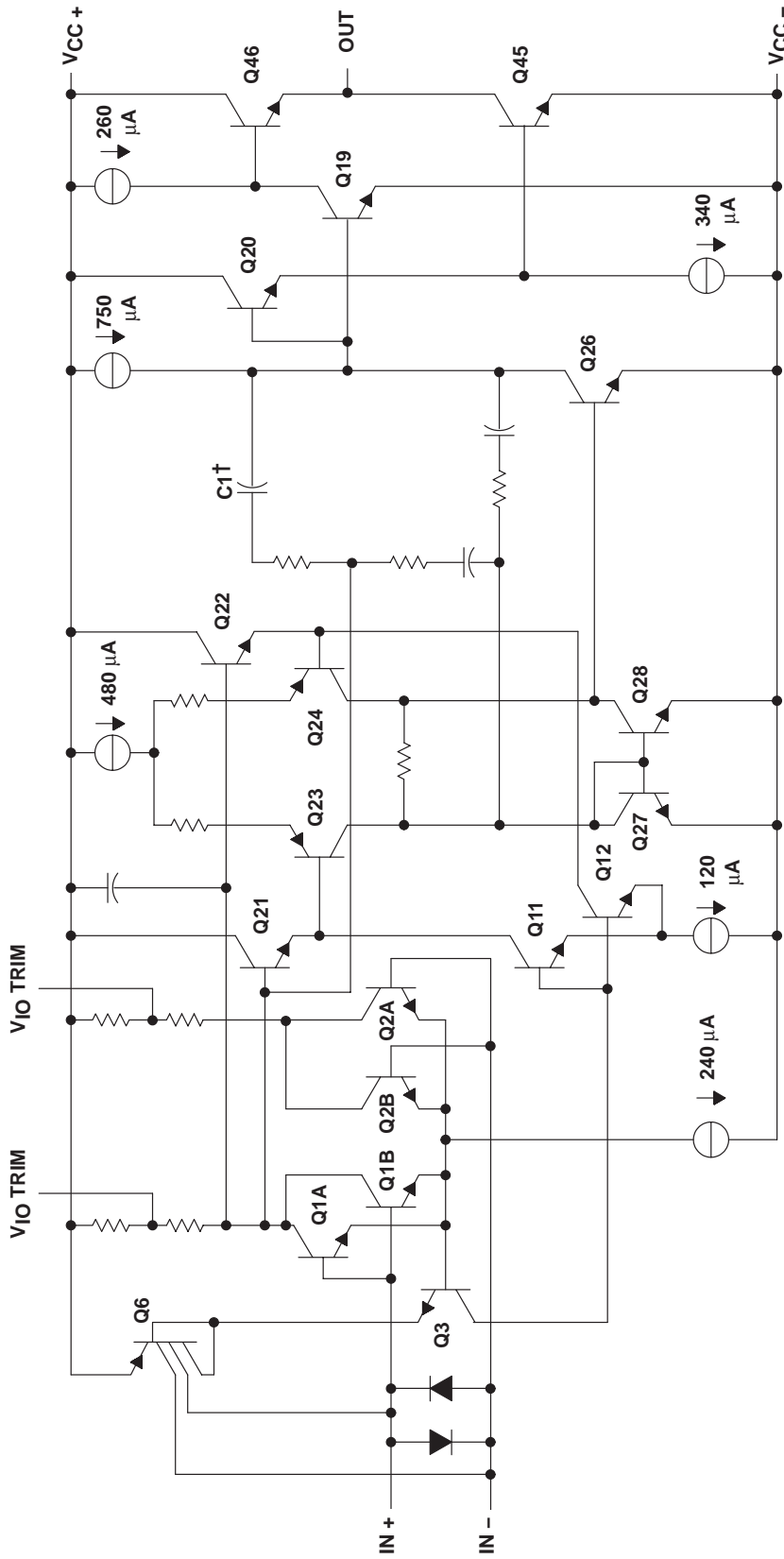
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# OP27A, OP27C LOW-NOISE HIGH-SPEED PRECISION OPERATIONAL-AMPLIFIER

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schematic



† C1 = 120 pF for OP27



# OP27A, OP27C

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**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{CC+}$ (see Note 1) .....	22 V
Supply voltage, $V_{CC-}$ (see Note 1) .....	22 V
Input voltage, $V_I$ .....	$V_{CC\pm}$
Duration of output short circuit .....	unlimited
Differential input current (see Note 2) .....	$\pm 25$ mA
Continuous power dissipation .....	See Dissipation Rating Table
Operating free-air temperature range: OP27A, OP27C .....	$-55^{\circ}\text{C}$ to $125^{\circ}\text{C}$
Storage temperature range .....	$-65^{\circ}\text{C}$ to $150^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or FK package .....	$300^{\circ}\text{C}$

- NOTES: 1. All voltage values are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$  unless otherwise noted.  
 2. The inputs are protected by back-to-back diodes. Current-limiting resistors are not used in order to achieve low noise. Excessive input current will flow if a differential input voltage in excess of approximately  $\pm 0.7$  V is applied between the inputs unless some limiting resistance is used.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
JG	1050 mW	8.4 mW/ $^{\circ}\text{C}$	546 mW	210 mW
FK	1375 mW	11.0 mW/ $^{\circ}\text{C}$	715 mW	275 mW



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## recommended operating conditions

		OP27A			OP27C			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
Supply voltage, $V_{CC+}$		4	15	22	4	15	22	V
Supply voltage, $V_{CC-}$		-4	-15	-22	-4	-15	-22	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} = \pm 15\text{ V}$ , $T_A = 25^\circ\text{C}$	$\pm 11$			$\pm 11$			V
	$V_{CC\pm} = \pm 15\text{ V}$ , $T_A = -55^\circ\text{C}$ to $125^\circ\text{C}$	$\pm 10.3$			$\pm 10.2$			
Operating free-air temperature, $T_A$		-55		125	-55		125	$^\circ\text{C}$

## electrical characteristics at specified free-air temperature, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	$T_A^\dagger$	OP27A			OP27C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	$V_O = 0$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ See Note 3	25 $^\circ\text{C}$	10		25	30		100	$\mu\text{V}$
			Full range	60			300			
$\alpha V_{IO}$	Average temperature coefficient of input offset voltage		Full range	0.2	0.6		0.4	1.8		$\mu\text{V}/^\circ\text{C}$
	Long-term drift of input offset voltage	See Note 4		0.2	1		0.4	2		$\mu\text{V}/\text{mo}$
$I_{IO}$	Input offset current	$V_O = 0$ , $V_{IC} = 0$	25 $^\circ\text{C}$	7	35		12	75		nA
			Full range	50			135			
$I_{IB}$	Input bias current	$V_O = 0$ , $V_{IC} = 0$	25 $^\circ\text{C}$	$\pm 10$	$\pm 40$		$\pm 15$	$\pm 80$		nA
			Full range	$\pm 60$			$\pm 150$			
$V_{ICR}$	Common-mode input voltage range		25 $^\circ\text{C}$	11 to -11			11 to -11			V
			Full range	10.3 to -10.3			10.5 to -10.5			
$V_{OM}$	Peak output voltage swing	$R_L \geq 2\ \text{k}\Omega$		$\pm 12$	$\pm 13.8$		$\pm 11.5$	$\pm 13.5$		V
				$\pm 10$	$\pm 11.5$		$\pm 10$	$\pm 11.5$		
			Full range	$\pm 11.5$			10.5			
$A_{VD}$	Large-signal differential voltage amplification	$R_L \geq 2\ \text{k}\Omega$ , $V_O = \pm 10\ \text{V}$		1000	1800		700	1500		V/mV
		$R_L \geq 1\ \text{k}\Omega$ , $V_O = \pm 10\ \text{V}$		800	1500		1500			
		$R_L \geq 0.6\ \text{k}\Omega$ , $V_O = \pm 1\ \text{V}$ , $V_{CC\pm} = \pm 4\ \text{V}$		250	700		200	500		
		$R_L \geq 2\ \text{k}\Omega$ , $V_O = \pm 10\ \text{V}$	Full range	600			300			
$r_{i(\text{CM})}$	Common-mode input resistance			3			2			G $\Omega$
$r_o$	Output resistance	$V_O = 0$ , $I_O = 0$	25 $^\circ\text{C}$	70			70			$\Omega$
CMRR	Common-mode rejection ratio	$V_{IC} = \pm 11\ \text{V}$	25 $^\circ\text{C}$	114	126		100	120		dB
		$V_{IC} = \pm 10\ \text{V}$	Full range	110			94			
$k_{\text{SVR}}$	Supply voltage rejection ratio	$V_{CC\pm} = \pm 4\ \text{V}$ to $\pm 18\ \text{V}$	25 $^\circ\text{C}$	100	120		94	118		dB
		$V_{CC\pm} = \pm 4.5\ \text{V}$ to $\pm 18\ \text{V}$	Full range	96			86			

$^\dagger$  Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

- NOTES: 3. Input offset voltage measurements are performed by automatic test equipment approximately 0.5 seconds after applying power.  
4. Long-term drift of input offset voltage refers to the average trend line of offset voltage versus time over extended periods after the first 30 days of operation. Excluding the initial hour of operation, changes in  $V_{IO}$  during the first 30 days are typically 2.5  $\mu\text{V}$  (see Figure 3).



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# OP27A, OP27C

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### OP27 operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	OP27A			OP27C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate	$A_{VD} \geq 1$ , $R_L \geq 2\text{ k}\Omega$	1.7	2.8		1.7	2.8		V/ $\mu$ s
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }10\text{ Hz}$ , $R_S = 20\ \Omega$ , See Figure 26		0.08	0.18		0.09	0.25	$\mu$ V
$V_n$	Equivalent input noise voltage	$f = 10\text{ Hz}$ , $R_S = 20\ \Omega$		3.5	5.5		3.8	8	nV/ $\sqrt{\text{Hz}}$
		$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$		3	3.8		3.2	4.5	
$I_n$	Equivalent input noise current	$f = 10\text{ Hz}$ , See Figure 27		5	15		5		pA/ $\sqrt{\text{Hz}}$
		$f = 1\text{ kHz}$ , See Figure 27		0.7	1.5		0.7	1.5	
	Gain-bandwidth product	$f = 100\text{ kHz}$	5	8		5	8		MHz

# OP27A, OP27C LOW-NOISE HIGH-SPEED PRECISION OPERATIONAL-AMPLIFIER

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## TYPICAL CHARACTERISTICS

Table of Graphs

		FIGURE
$V_{IO}$	Input offset voltage	vs Temperature 1
$\Delta V_{IO}$	Change in input offset voltage	vs Time after power on vs Time (long-term drift) 2 3
$I_{IO}$	Input offset current	vs Temperature 4
$I_{IB}$	Input bias current	vs Temperature 5
$V_{ICR}$	Common-mode input voltage range	vs Supply voltage 6
$V_{OM}$	Maximum peak output voltage	vs Load resistance 7
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency 8
$A_{VD}$	Differential voltage amplification	vs Supply voltage vs Load resistance vs Frequency 9 10 11, 12
CMRR	Common-mode rejection ratio	vs Frequency 13
$k_{SVR}$	Supply voltage rejection ratio	vs Frequency 14
SR	Slew rate	vs Temperature 15
$\phi_m$	Phase margin	vs Temperature 16
$\phi$	Phase shift	vs Frequency 11
$V_n$	Equivalent input noise voltage	vs Bandwidth vs Source resistance vs Supply voltage vs Temperature vs Frequency 17 18 19 20 21
	Gain-bandwidth product	vs Temperature 16
$I_{OS}$	Short-circuit output current	vs Time 22
$I_{CC}$	Supply current	vs Supply voltage 23
	Pulse response	Small signal Large signal 24 25



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# OP27A, OP27C

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### TYPICAL CHARACTERISTICS

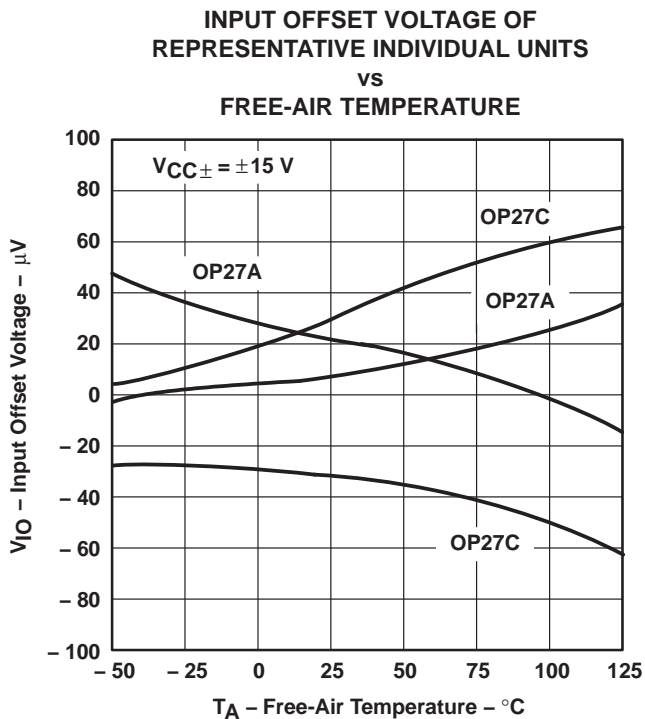


Figure 1

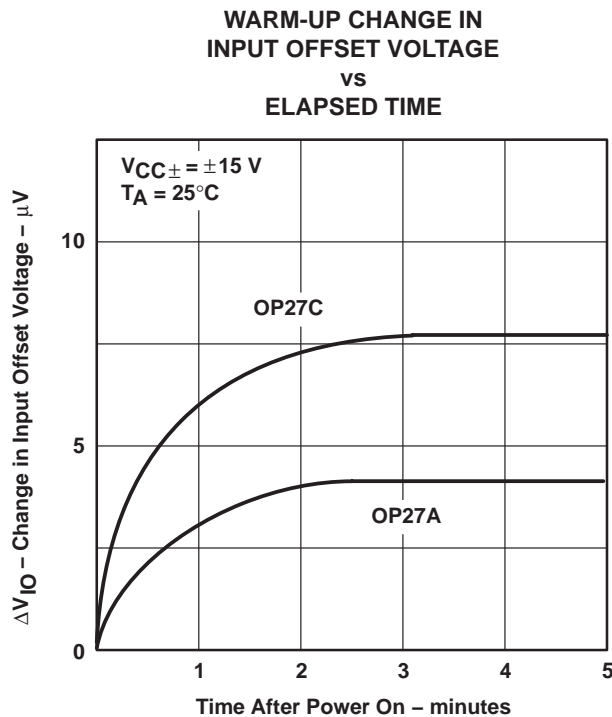


Figure 2

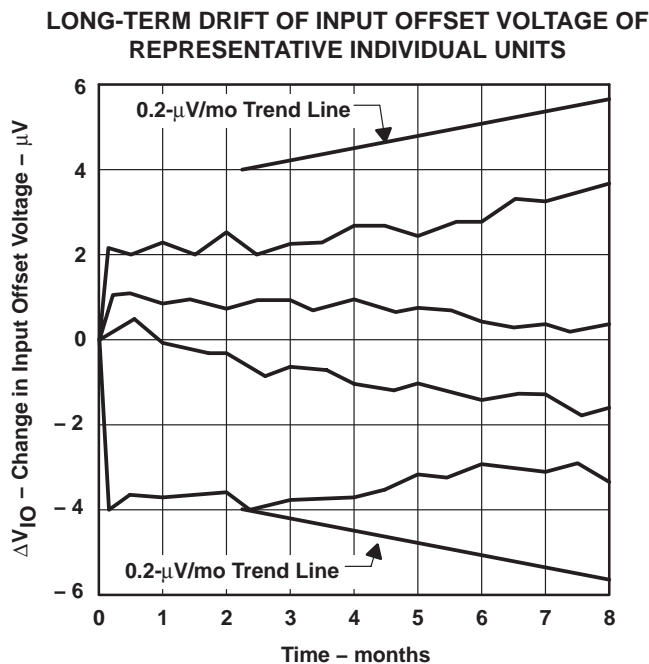


Figure 3



# OP27A, OP27C LOW-NOISE HIGH-SPEED PRECISION OPERATIONAL-AMPLIFIER

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## TYPICAL CHARACTERISTICS

**INPUT OFFSET CURRENT  
vs  
FREE-AIR TEMPERATURE**

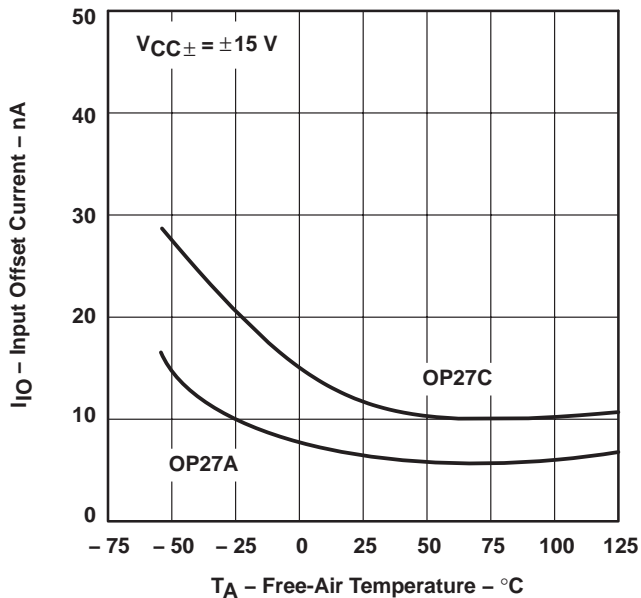


Figure 4

**INPUT BIAS CURRENT  
vs  
FREE-AIR TEMPERATURE**

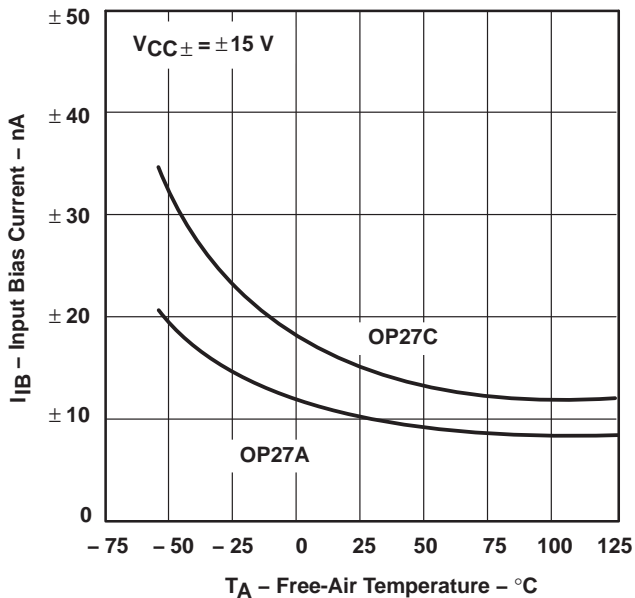


Figure 5

**COMMON-MODE INPUT VOLTAGE RANGE LIMITS  
vs  
SUPPLY VOLTAGE**

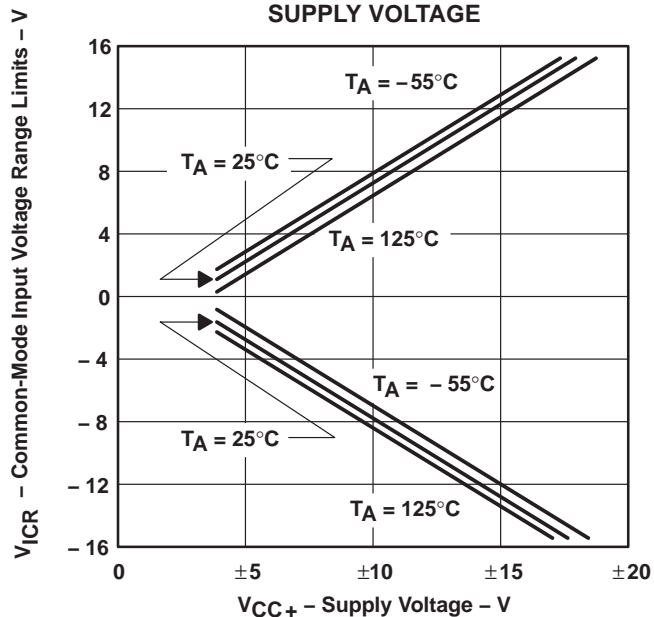


Figure 6

**MAXIMUM PEAK OUTPUT VOLTAGE  
vs  
LOAD RESISTANCE**

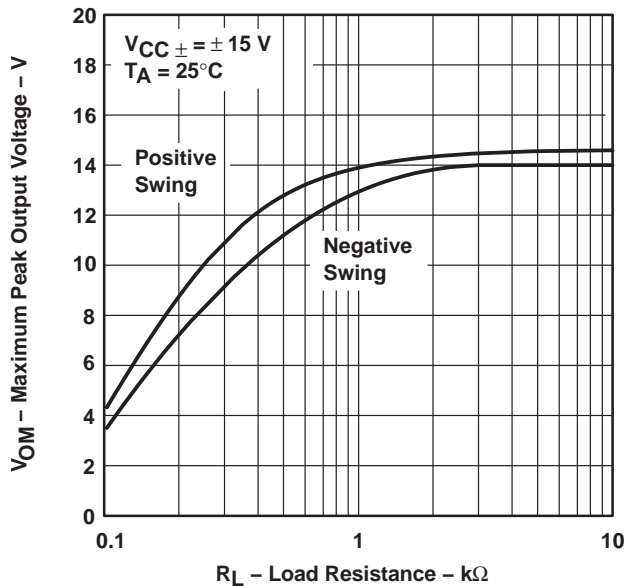


Figure 7





**TYPICAL CHARACTERISTICS**

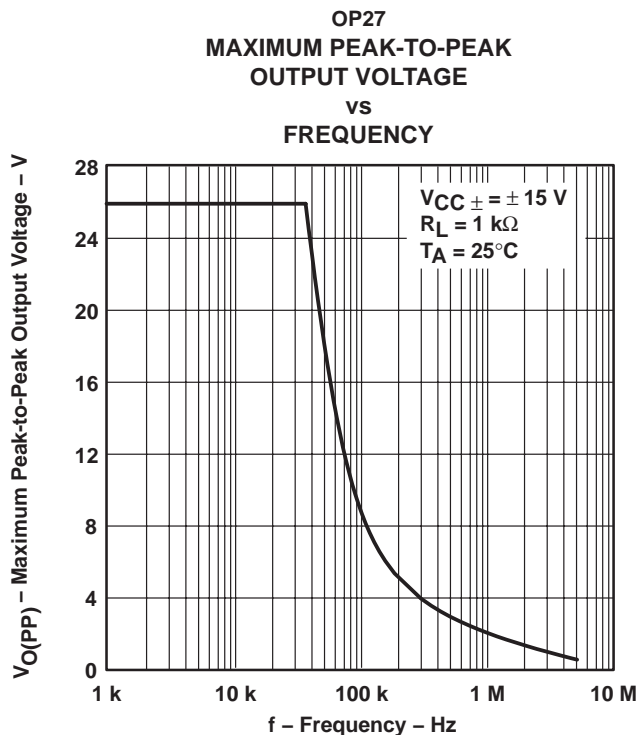


Figure 8.

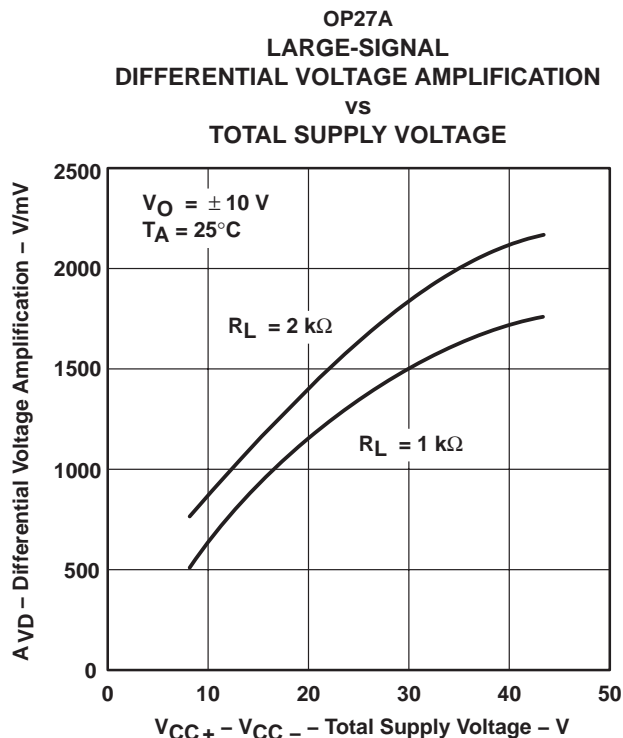


Figure 9

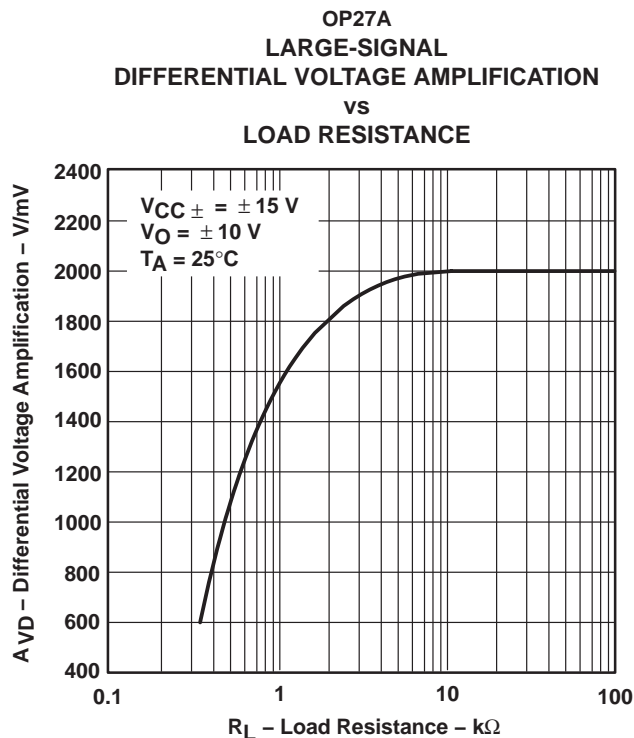


Figure 10

# OP27A, OP27C LOW-NOISE HIGH-SPEED PRECISION OPERATIONAL-AMPLIFIER

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## TYPICAL CHARACTERISTICS

OP27  
LARGE-SIGNAL DIFFERENTIAL  
VOLTAGE AMPLIFICATION AND PHASE SHIFT  
VS  
FREQUENCY

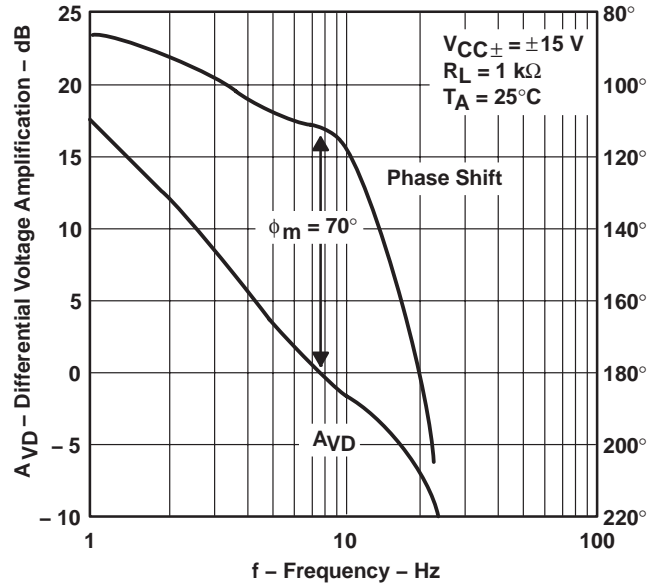


Figure 11.

OP27A  
LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
VS  
FREQUENCY

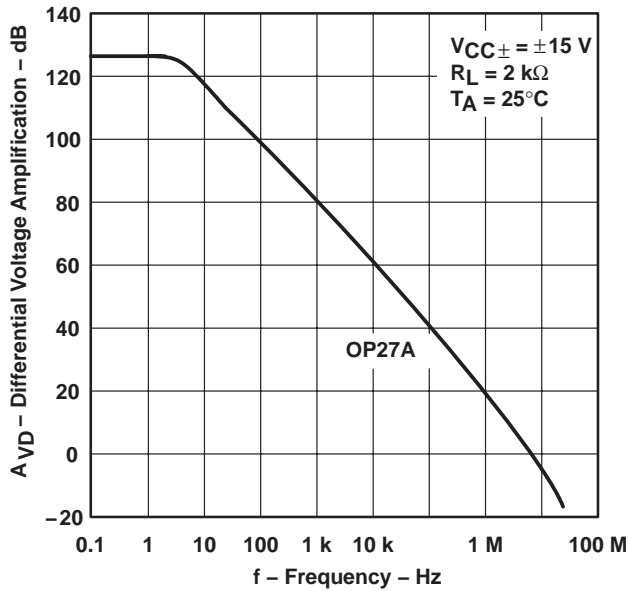


Figure 12

OP27A  
COMMON-MODE REJECTION RATIO  
VS  
FREQUENCY

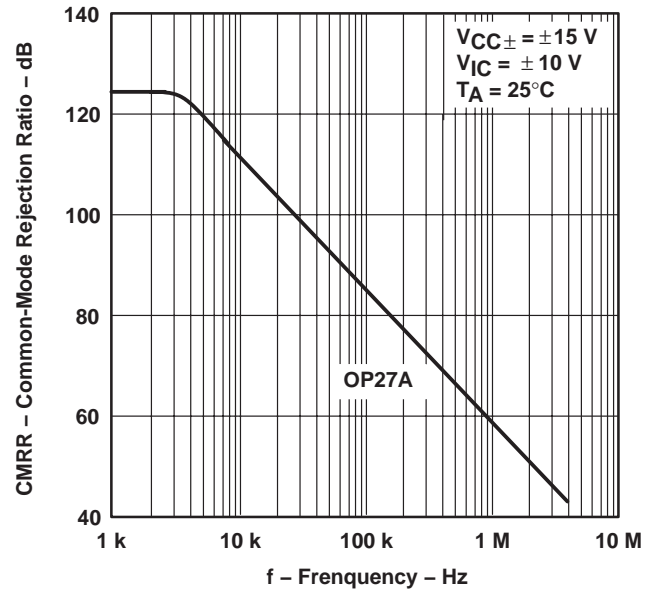
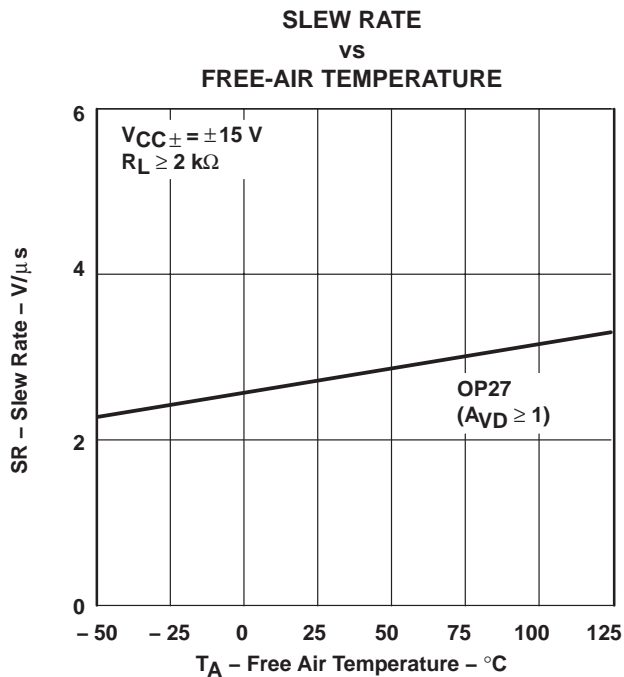
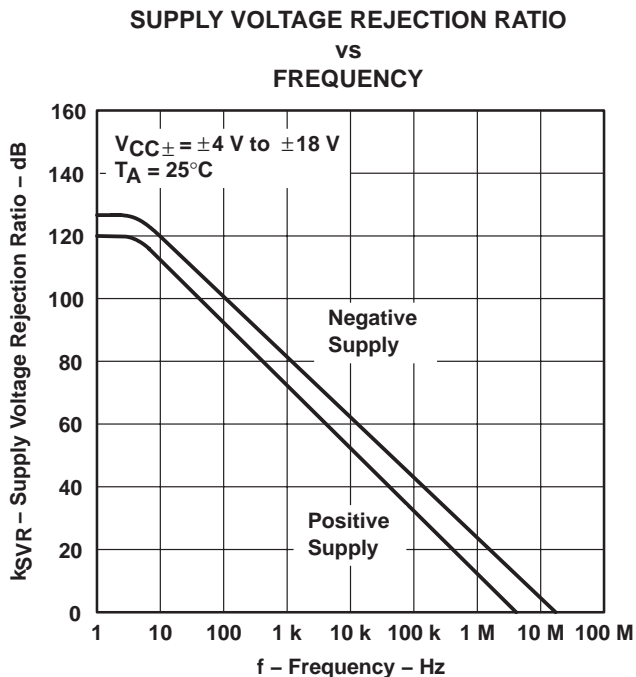


Figure 13

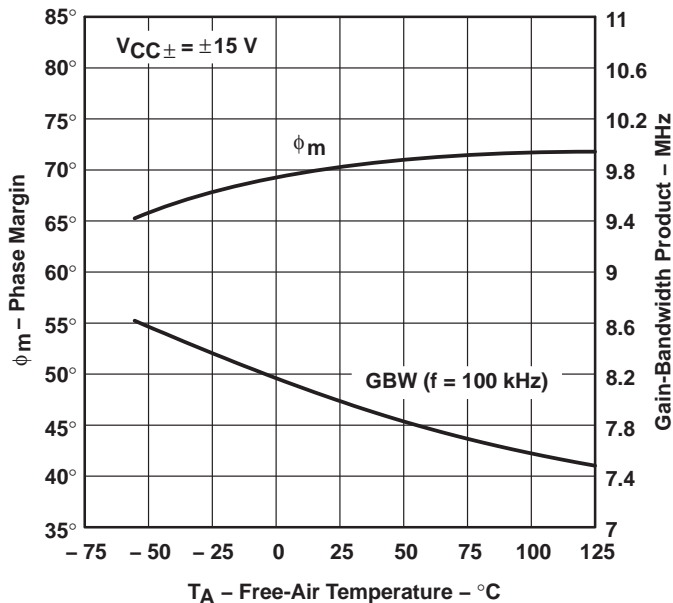


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**TYPICAL CHARACTERISTICS**



**OP27  
PHASE MARGIN AND  
GAIN-BANDWIDTH PRODUCT  
vs  
FREE-AIR TEMPERATURE**



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## TYPICAL CHARACTERISTICS

EQUIVALENT INPUT NOISE VOLTAGE  
vs  
BANDWIDTH

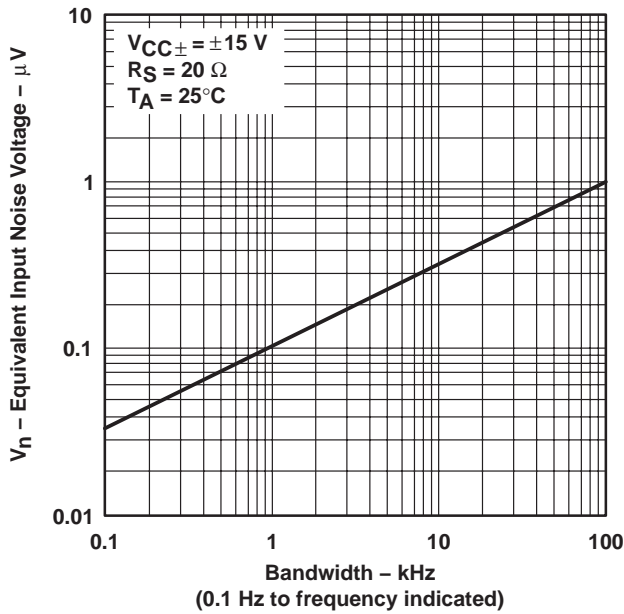


Figure 17

OP27A  
EQUIVALENT INPUT NOISE VOLTAGE  
vs  
TOTAL SUPPLY VOLTAGE

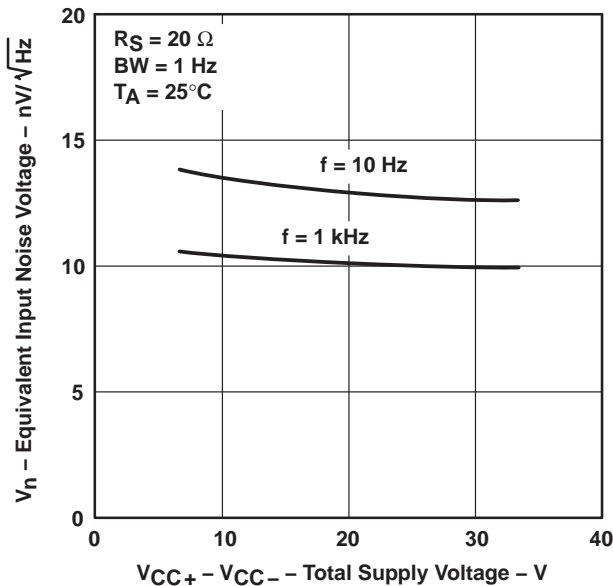


Figure 19

TOTAL EQUIVALENT INPUT NOISE VOLTAGE  
vs  
SOURCE RESISTANCE

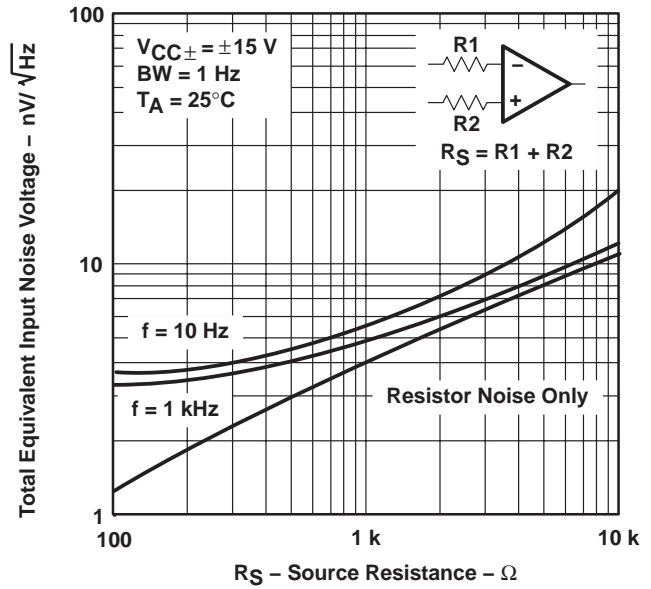


Figure 18

OP27A  
EQUIVALENT INPUT NOISE VOLTAGE  
vs  
FREE-AIR TEMPERATURE

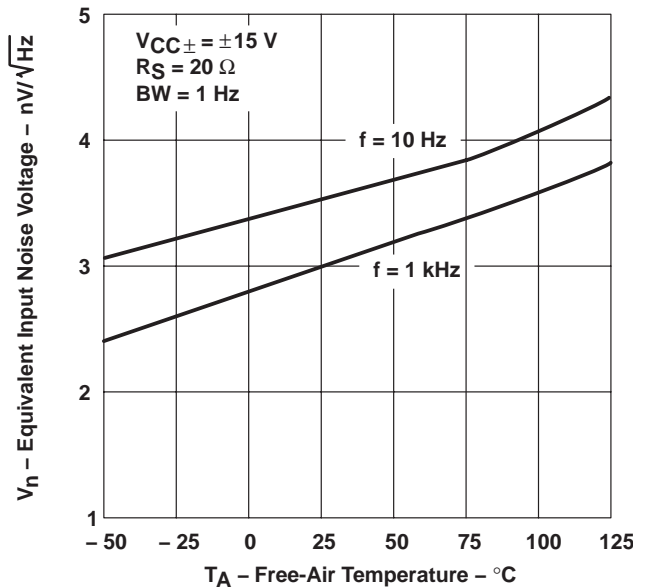
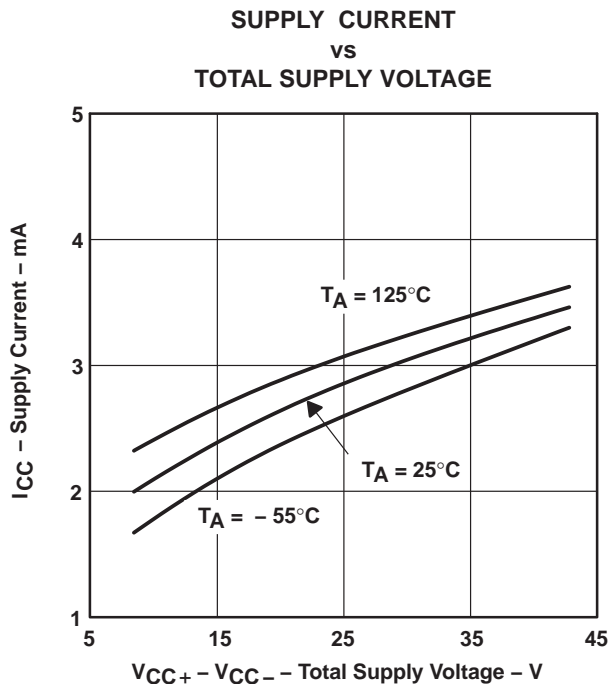
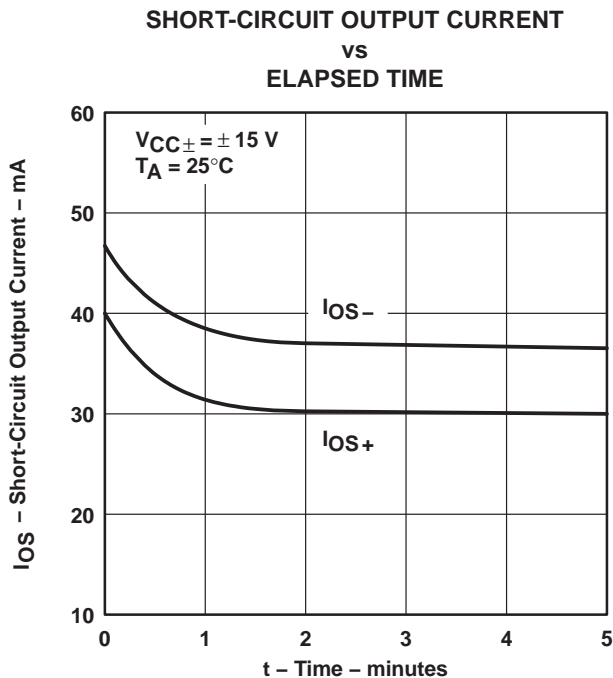
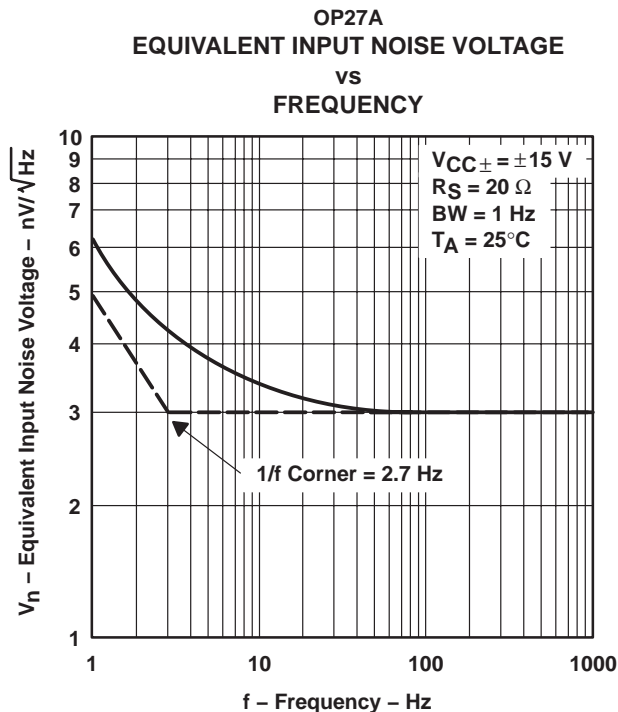


Figure 20



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**TYPICAL CHARACTERISTICS**



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## TYPICAL CHARACTERISTICS

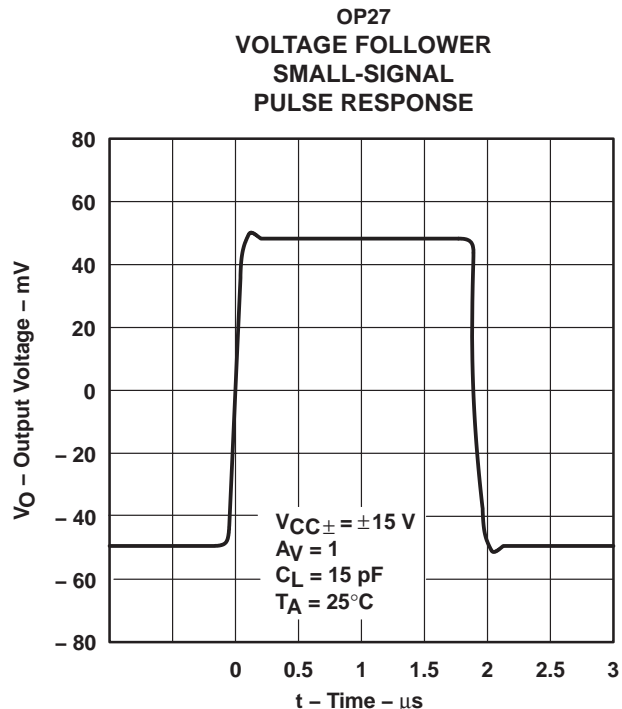


Figure 24

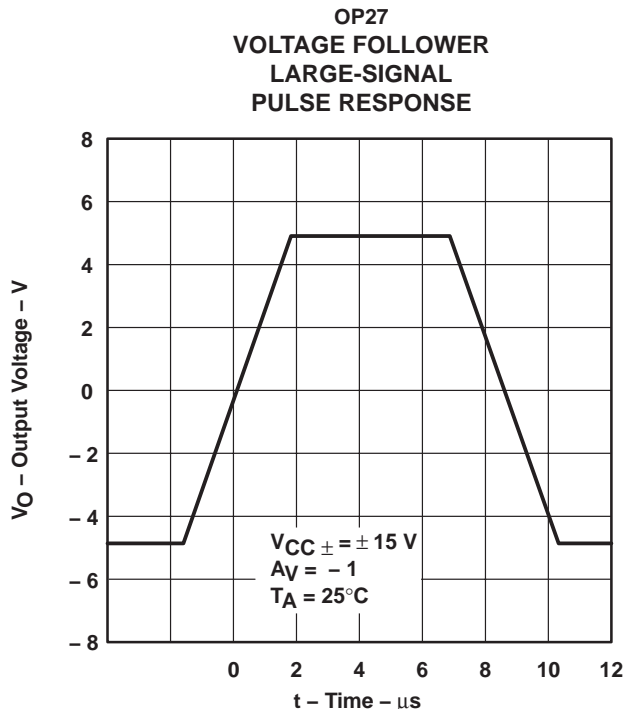


Figure 25

## APPLICATION INFORMATION

### general

The OP27 series devices can be inserted directly onto OP07, OP05,  $\mu\text{A}725$ , and SE5534 sockets with or without removing external compensation or nulling components. In addition, the OP27 can be fitted to  $\mu\text{A}741$  sockets by removing or modifying external nulling components.

### noise testing

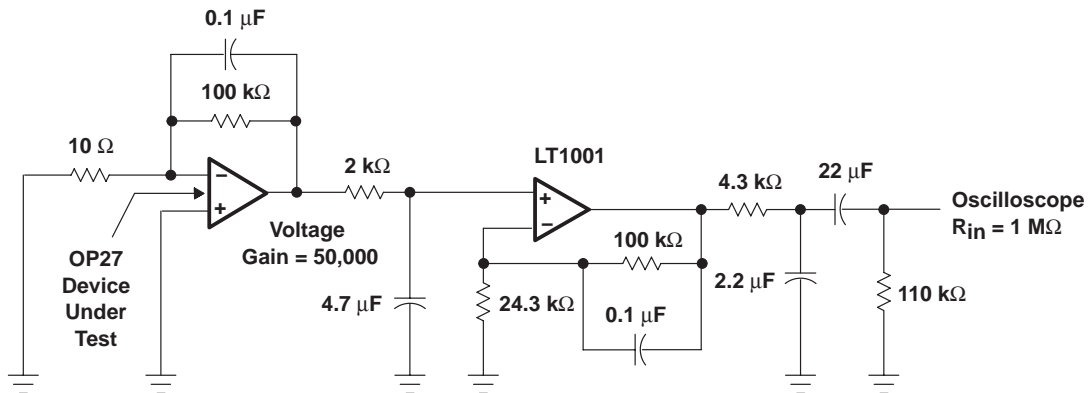
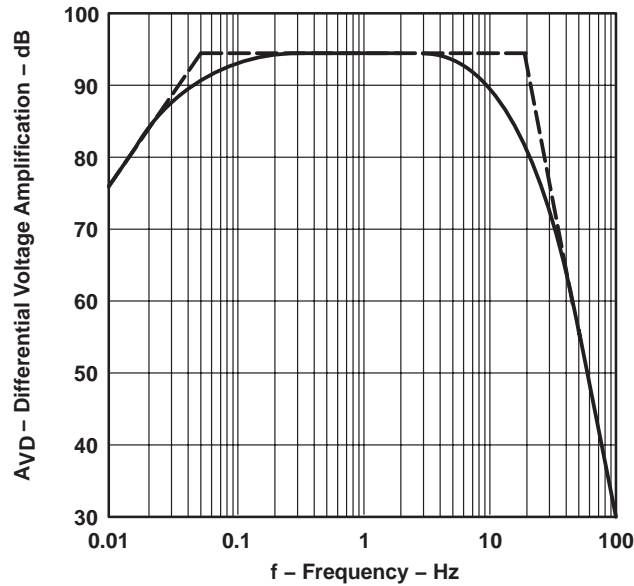
Figure 26 shows a test circuit for 0.1-Hz to 10-Hz peak-to-peak noise measurement of the OP27. The frequency response of this noise tester indicates that the 0.1-Hz corner is defined by only one zero. Because the time limit acts as an additional zero to eliminate noise contributions from the frequency band below 0.1 Hz, the test time to measure 0.1-Hz to 10-Hz noise should not exceed 10 seconds.

Measuring the typical 80-nV peak-to-peak noise performance of the OP27 requires the following special test precautions:

**APPLICATION INFORMATION**

**noise testing (continued)**

1. The device should be warmed up for at least five minutes. As the operational amplifier warms up, the offset voltage typically changes  $4 \mu\text{V}$  due to the chip temperature increasing from  $10^\circ\text{C}$  to  $20^\circ\text{C}$  starting from the moment the power supplies are turned on. In the 10-s measurement interval, these temperature-induced effects can easily exceed tens of nanovolts.
2. For similar reasons, the device should be well shielded from air currents to eliminate the possibility of thermoelectric effects in excess of a few nanovolts, which would invalidate the measurements.
3. Sudden motion in the vicinity of the device should be avoided, as it produces a feedthrough effect that increases observed noise.



NOTE: All capacitor values are for nonpolarized capacitors only.

**Figure 26. 0.1-Hz to 10-Hz Peak-to-Peak Noise Test Circuit and Frequency Response**

# OP27A, OP27C LOW-NOISE HIGH-SPEED PRECISION OPERATIONAL-AMPLIFIER

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## APPLICATION INFORMATION

### noise testing (continued)

When measuring noise on a large number of units, a noise-voltage density test is recommended. A 10-Hz noise-voltage density measurement correlates well with a 0.1-Hz to 10-Hz peak-to-peak noise reading since both results are determined by the white noise and the location of the 1/f corner frequency.

Figure 27 shows a circuit measuring current noise and the formula for calculating current noise.

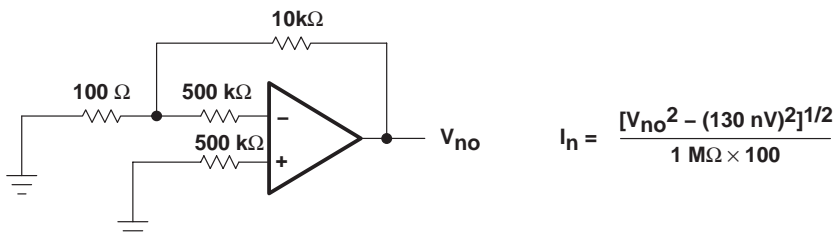


Figure 27. Current Noise Test Circuit and Formula

### offset voltage adjustment

The input offset voltage and temperature coefficient of the OP27 are permanently trimmed to a low level at wafer testing. However, if further adjustment of  $V_{IO}$  is necessary, using a 10-kΩ nulling potentiometer as shown in Figure 28 does not degrade the temperature coefficient  $\alpha_{V_{IO}}$ . Trimming to a value other than zero creates an  $\alpha_{V_{IO}}$  of  $V_{IO}/300 \mu\text{V}/^\circ\text{C}$ . For example, if  $V_{IO}$  is adjusted to 300  $\mu\text{V}$ , the change in  $\alpha_{V_{IO}}$  is 1  $\mu\text{V}/^\circ\text{C}$ .

The adjustment range with a 10-kΩ potentiometer is approximately  $\pm 2.5 \text{ mV}$ . If a smaller adjustment range is needed, the sensitivity and resolution of the nulling can be improved by using a smaller potentiometer in conjunction with fixed resistors. The example in Figure 29 has an approximate null range of  $\pm 200 \mu\text{V}$ .

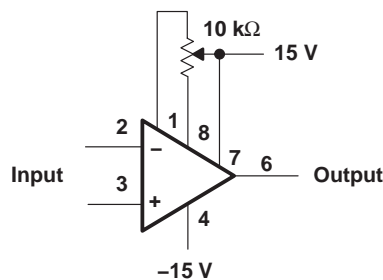


Figure 28. Standard Input Offset Voltage Adjustment

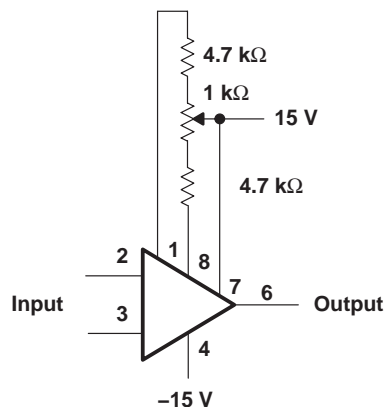


Figure 29. Input Offset Voltage Adjustment With Improved Sensitivity

### offset voltage and drift

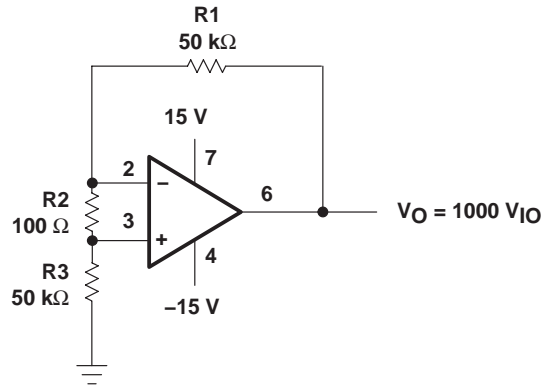
Unless proper care is exercised, thermoelectric effects caused by temperature gradients across dissimilar metals at the contacts to the input terminals can exceed the inherent temperature coefficient  $\alpha_{V_{IO}}$  of the amplifier. Air currents should be minimized, package leads should be short, and the two input leads should be close together and at the same temperature.



**APPLICATION INFORMATION**

**offset voltage and drift (continued)**

The circuit shown in Figure 30 measures offset voltage. This circuit can also be used as the burn-in configuration for the OP27 with the supply voltage increased to 20 V,  $R_1 = R_3 = 10\text{ k}\Omega$ ,  $R_2 = 200\ \Omega$ , and  $A_{VD} = 100$ .



NOTE A: Resistors must have low thermoelectric potential.

**Figure 30. Test Circuit for Offset Voltage and Offset Voltage Temperature Coefficient**

**unity gain buffer applications**

The resulting output waveform, when  $R_f \leq 100\ \Omega$  and the input is driven with a fast large-signal pulse ( $>1\text{ V}$ ), is shown in the pulsed-operation diagram in Figure 31.



**Figure 31. Pulsed Operation**

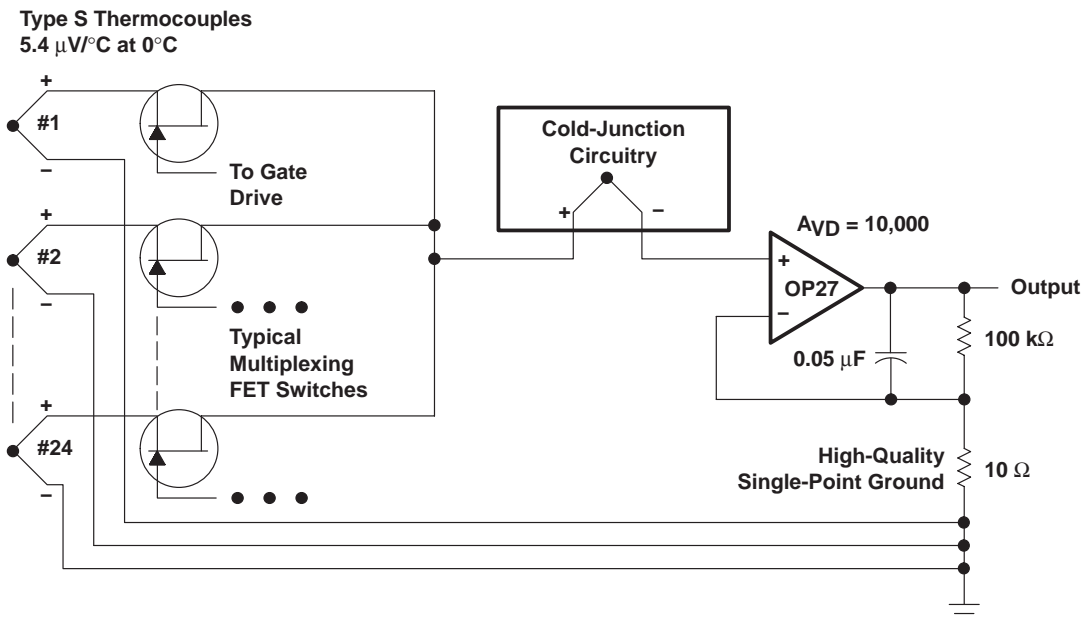
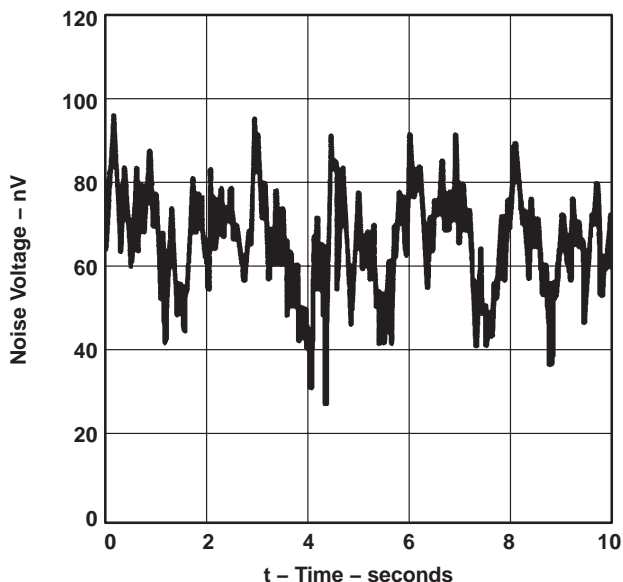
During the initial (fast-feedthrough-like) portion of the output waveform, the input protection diodes effectively short the output to the input, and a current, limited only by the output short-circuit protection, is drawn by the signal generator. When  $R_f \geq 500\ \Omega$ , the output is capable of handling the current requirements (load current  $\leq 20\text{ mA}$  at  $10\text{ V}$ ), the amplifier stays in its active mode, and a smooth transition occurs. When  $R_f > 2\text{ k}\Omega$ , a pole is created with  $R_f$  and the amplifier's input capacitance, creating additional phase shift and reducing the phase margin. A small capacitor ( $20\text{ pF}$  to  $50\text{ pF}$ ) in parallel with  $R_f$  eliminates this problem.

# OP27A, OP27C LOW-NOISE HIGH-SPEED PRECISION OPERATIONAL-AMPLIFIER

SLOS100D – FEBRUARY 1989 – REVISED JANUARY 2007

## APPLICATION INFORMATION

### unity gain buffer applications (continued)



NOTE A: If 24 channels are multiplexed per second and the output is required to settle to 0.1 % accuracy, the amplifier's bandwidth cannot be limited to less than 30 Hz. The peak-to-peak noise contribution of the OP27 will still be only 0.11  $\mu\text{V}$ , which is equivalent to an error of only  $0.02^\circ\text{C}$ .

**Figure 32. Low-Noise, Multiplexed Thermocouple Amplifier and 0.1-Hz to 10-Hz Peak-to-Peak Noise Voltage**

**PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
JM38510/13503BPA	OBSOLETE	CDIP	JG	8		TBD	Call TI	Call TI
JM38510/13506BPA	ACTIVE	CDIP	JG	8	1	TBD	A42 SNPB	N / A for Pkg Type
OP27AFKB	ACTIVE	LCCC	FK	20	1	TBD	POST-PLATE	N / A for Pkg Type
OP27AJGB	ACTIVE	CDIP	JG	8	1	TBD	A42 SNPB	N / A for Pkg Type
OP27CJGB	ACTIVE	CDIP	JG	8	1	TBD	A42 SNPB	N / A for Pkg Type

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

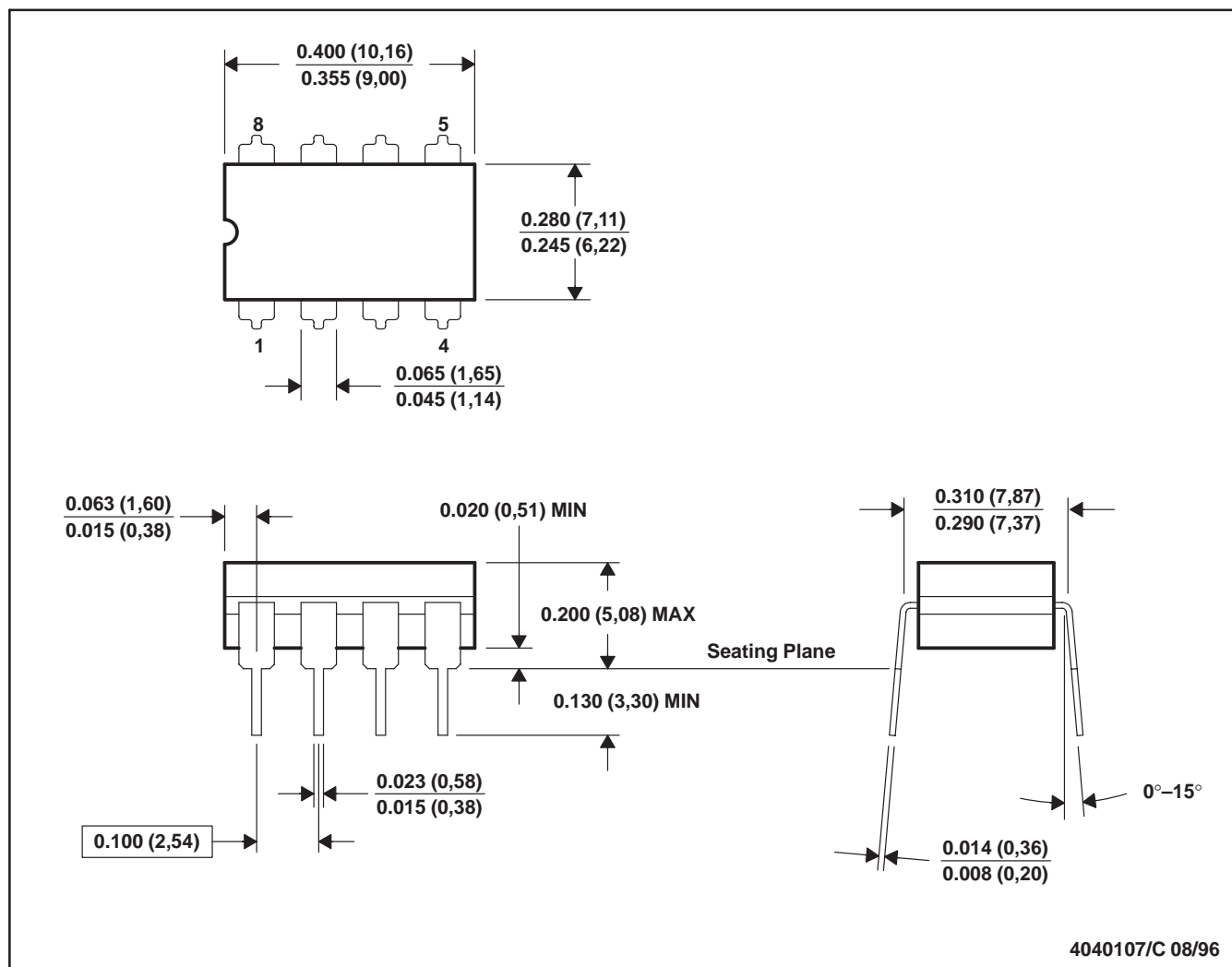
<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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JG (R-GDIP-T8)

CERAMIC DUAL-IN-LINE

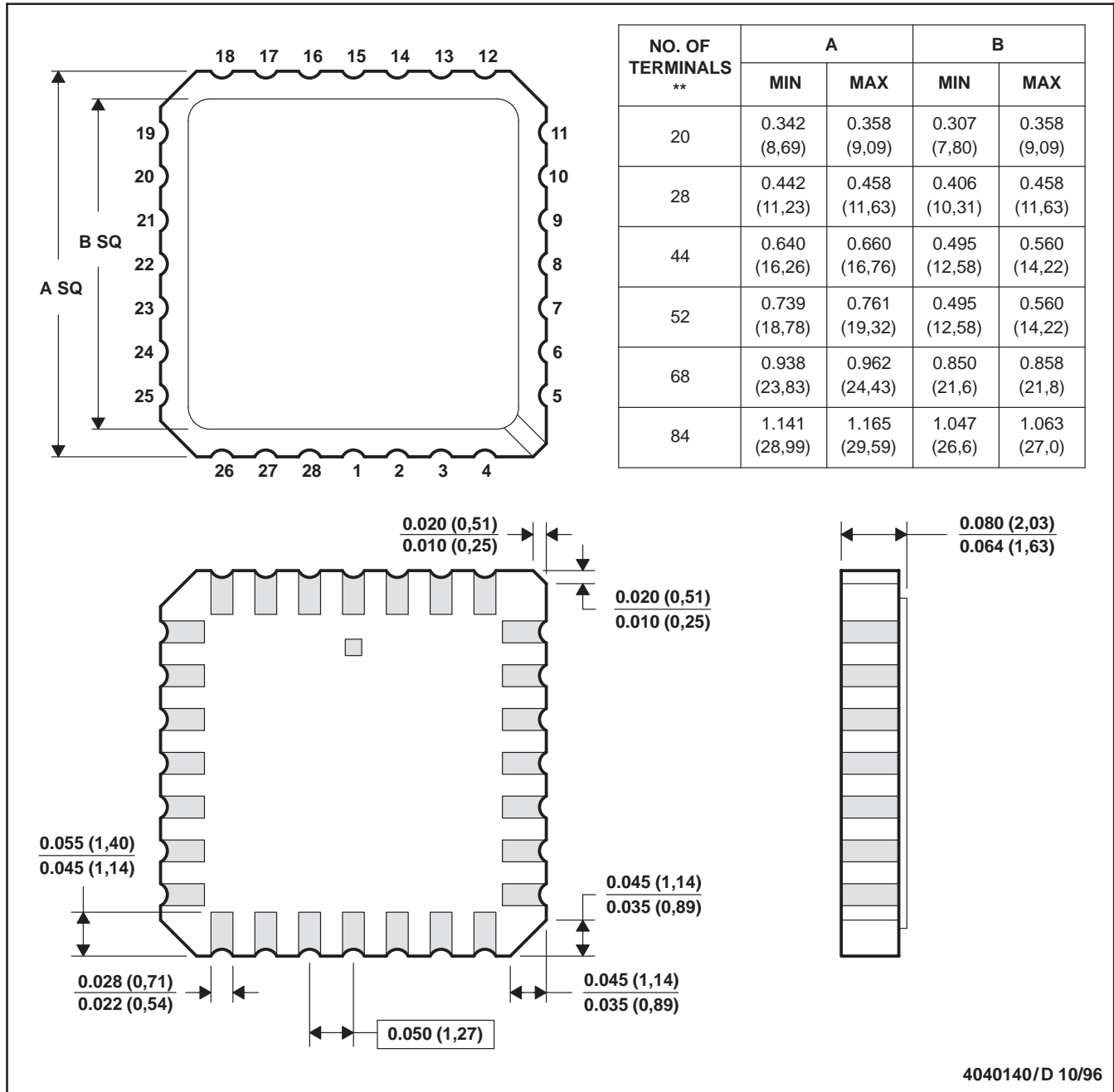


- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. This package can be hermetically sealed with a ceramic lid using glass frit.  
 D. Index point is provided on cap for terminal identification.  
 E. Falls within MIL STD 1835 GDIP1-T8

FK (S-CQCC-N\*\*)

LEADLESS CERAMIC CHIP CARRIER

28 TERMINAL SHOWN



- NOTES:
- A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  - C. This package can be hermetically sealed with a metal lid.
  - D. The terminals are gold plated.
  - E. Falls within JEDEC MS-004

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