DUAL DIFFERENTIAL PECL DRIVER/RECEIVER

FEATURES

- **Functional Replacement for the Agere BTF1A**
- **Driver Features**
  - **Third-State Logic Low Output**
  - **ESD Protection HBM > 3 kV, CDM > 2 kV**
  - **No Line Loading when V_{CC} = 0**
  - **Capable of Driving 50-Ω loads**
  - **2.0-ns Maximum Propagation Delay**
  - **0.2-ns Output Skew (typical)**
- **Receiver Features**
  - **High-Input Impedance Approximately 8 kΩ**
  - **4.0-ns Maximum Propagation Delay**
  - **50-mV Hysteresis**
  - **Slew Rate Limited (1 ns min 80% to 20%)**
  - **ESD Protection HBM > 3 kV, CDM > 2 kV**
  - **-1.1-V to 7.1-V Input Voltage Range**
- **Common Device Features**
  - **Common Enable for Each Driver/Receiver Pair**
  - **Operating Temperature Range: -40°C to 85°C**
  - **Single 5.0 V ± 10% Supply**
  - **Available in Gull-Wing SOIC (JEDEC MS-013, DW) and SOIC (D) Package**

DESCRIPTION

The TB5T1 device is a dual differential driver/receiver circuit that transmits and receives digital data over balanced transmission lines. The dual drivers translate input TTL logic levels to differential pseudo-ECL output levels. The dual receivers convert differential-input logic levels to TTL output levels. Each driver or receiver pair has its own common enable control allowing serial data and a control clock to be transmitted and received on a single integrated circuit. The TB5T1 requires the customer to supply termination resistors on the circuit board.
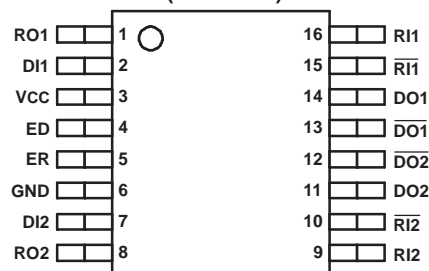
The power-down loading characteristics of the receiver input circuit are approximately 8 kΩ relative to the power supplies; hence, it does not load the transmission line when the circuit is powered down.

In circuits with termination resistors, the line remains impedance- matched when the circuit is powered down. The driver does not load the line when it is powered down.
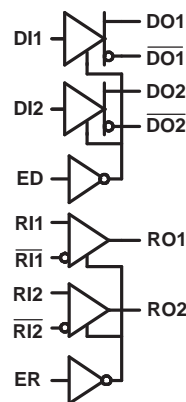
All devices are characterized for operation from -40°C to 85°C.

The logic inputs of this device include internal pull-up resistors of approximately 40 kΩ that are connected to V_{CC} to ensure a logical high level input if the inputs are open circuited.

PIN ASSIGNMENTS
DW AND D PACKAGE
(TOP VIEW)

RO1	1	16	RI1
DI1	2	15	$\overline{RI1}$
VCC	3	14	DO1
ED	4	13	$\overline{DO1}$
ER	5	12	$\overline{DO2}$
GND	6	11	DO2
DI2	7	10	$\overline{RI2}$
RO2	8	9	RI2

FUNCTIONAL BLOCK DIAGRAM

Enable Truth Table

ED	ER	D1	D2	R1	R2
0	0	Active	Active	Active	Active
1	0	Disabled	Disabled	Active	Active
0	1	Active	Active	Disabled	Disabled
1	1	Disabled	Disabled	Disabled	Disabled

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ORDERING INFORMATION

PART NUMBER	PART MARKING	PACKAGE[1]	LEAD FINISH	STATUS
TB5T1DW	TB5T1	Gull-Wing SOIC	NiPdAu	Production
TB5T1D	TB5T1	SOIC	NiPdAu	Production

(1)   For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

POWER DISSIPATION RATINGS

PACKAGE	CIRCUIT BOARD MODEL	POWER RATING $T_A \leq 25°C$	THERMAL RESISTANCE, JUNCTION-TO-AMBIENT WITH NO AIR FLOW	DERATING FACTOR [1] $T_A \geq 25°C$	POWER RATING $T_A = 85°C$
D	Low-K[2]	752 mW	132.8°C/W	7.5 mW/°C	301 mW
	High-K[3]	1160 mW	85.8°C/W	11.7 mW/°C	466 mW
DW	Low-K[2]	814 mW	122.7°C/W	8.2 mW/°C	325 mW
	High-K[3]	1200 mW	83.1°C/W	12 mW/°C	481 mW

(1)   This is the inverse of the junction-to-ambient thermal resistance when board-mounted with no air flow.
(2)   In accordance with the low-K thermal metric definitions of EIA/JESD51-3.
(3)   In accordance with the high-K thermal metric definitions of EIA/JESD51-7.

THERMAL CHARACTERISTICS

PARAMETER		PACKAGE	VALUE	UNIT
θ_{JB}	Junction-to-board thermal resistance	D	48.4	°C/W
		DW	55.2	°C/W
θ_{JC}	Junction-to-case thermal resistance	D	45.1	°C/W
		DW	48.1	°C/W

Product Folder Link(s): *TB5T1*

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range unless otherwise noted[1]

			UNIT
Supply voltage, V_{CC}			0 V to 6 V
Magnitude of differential bus (input) voltage, $|V_{RI1} - V_{\overline{RI1}}|$, $|V_{RI2} - V_{\overline{RI2}}|$			8.4 V
ESD	Human Body Model [2]	All pins	±3 kV
	Charged-Device Model [3]	All pins	±2 kV
Continuous power dissipation			See Dissipation Rating Table
Storage temperature, T_{stg}			-65°C to 150°C

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) Tested in accordance with JEDEC Standard 22, Test Method A114-A.
(3) Tested in accordance with JEDEC Standard 22, Test Method C101.

RECOMMENDED OPERATING CONDITIONS

	MIN	NOM	MAX	UNIT
Supply voltage, V_{CC}	4.5	5	5.5	V
Bus pin input voltage, V_{RI1}, $V_{\overline{RI1}}$, V_{RI2}, or $V_{\overline{RI2}}$	-1.2[1]		7.2	V
Magnitude of differential input voltage, $|V_{RI1} - V_{\overline{RI1}}|$, $|V_{RI2} - V_{\overline{RI2}}|$	0.1		6	V
Operating free-air temperature, T_A	-40		85	°C

(1) The algebraic convention, in which the least positive (most negative) limit is designated as minimum is used in this data sheet, unless otherwise noted.

ELECTRICAL CHARACTERISTICS

over operating free-air temperature range unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{CC}	Supply current	Outputs disabled			40	mA
		Outputs enabled			40	mA

THIRD STATE

A TB5T1 driver produces pseudo-ECL levels and has a third state mode, which is different from a conventional TTL device. When a TB5T1 driver is placed in the third state, the base of the output transistors are pulled low, bringing the outputs below the active-low level of standard PECL devices. (For example: The TB5T1 low output level is typically 2.7 V, while the third state noninverting output level is typically 1.2 V.) In a bidirectional, multipoint bus application, the driver of one device, which is in its third state, can be back driven by another driver on the bus whose voltage in the low state is lower than the 3-stated device. This could be due to differences between individual driver's power supplies. In this case, the device in the third state controls the line, thus clamping the line and reducing the signal swing. If the difference between the driver power supplies is small, this consideration can be ignored. Again using the TB5T1 driver as an example, a typical supply voltage difference between separate drivers of > 2 V can exist without significantly affecting the amplitude of the signal.

DRIVER ELECTRICAL CHARACTERISTICS

over operating free-air temperature range unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OH}	Output high voltage[1]		V_{CC} - 1.8	V_{CC} - 1.3	V_{CC} - 0.8	V
V_{OL}	Output low voltage[1]		V_{OH} - 1.4	V_{OH} - 1.2	V_{OH} - 0.7	V
V_{OD}	Differential output voltage, $|V_{OH} - V_{OL}|$		0.7	1.1	1.4	V
V_{OH}	Output high voltage[1]		V_{CC} - 1.8	V_{CC} - 1.3	V_{CC} - 0.8	V
V_{OL}	Output low voltage[1]	T_A = 0C to 85C	V_{OH} - 1.4	V_{OH} - 1.1	V_{OH} - 0.5	V
V_{OD}	Differential output voltage, $|V_{OH} - V_{OL}|$		0.5	1.1	1.4	V
$V_{OC(PP)}$	Peak-to-peak common-mode output voltage	C_L= 5 pF, See Figure 7		230	600	mV
V_{OZH}	Third state output high voltage[1] $\overline{DO1}, \overline{DO2}$	V_{CC} = 4.5 V	1.4	1.8	2.2	V
V_{OZD}	Third state deferential output voltage[1] $V_{DOn} - V_{\overline{DOn}}$		-0.47[2]	-0.6		
V_{IL}	Input low voltage[3]	V_{CC} = 5.5 V			0.8	V
V_{IH}	Input high voltage	V_{CC} = 4.5 V	2			V
V_{IK}	Input clamp voltage	V_{CC} = 4.5 V, I_I = -5 mA			-1[2]	V
I_{OS}	Short-circuit output current[4]	V_{CC} = 5.5 V, V_O = 0 V			-250[2]	mA
		V_{CC} = 5.5 V, V_{OD} = 0 V			10[2]	mA
I_{IL}	Input low current	V_{CC} = 5.5 V, V_I = 0.4 V			-400[2]	A
I_{IH}	Input high current	V_{CC} = 5.5 V, V_I = 2.7 V			20	A
I_{IH}	Input reverse current	V_{CC} = 5.5 V, V_I = 5.5 V			100	A
C_{IN}	Input Capacitance			5		pF

(1) Values are with terminations as per Figure 6.
(2) This parameter is listed using a magnitude and polarity/direction convention, rather than an algebraic convention, to match the original Agere data sheet.
(3) The input levels and difference voltage provide no noise immunity and should be tested only in a static, noise-free environment.
(4) Test must be performed one lead at a time to prevent damage to the device. No test circuit attached.

RECEIVER ELECTRICAL CHARACTERISTICS

over operating free-air temperature range unless otherwise noted

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
V_{OL}	Output low voltage	V_{CC} = 4.5 V, I_{OL} = 8.0 mA				0.4	V
V_{OH}	Output high voltage	V_{CC} = 4.5 V, I_{OH} = -400 µA		2.4			V
V_{IL}	Enable input low voltage[1]	V_{CC}= 5.5 V				0.8	V
V_{IH}	Enable input high voltage[1]	V_{CC} = 4.5 V		2			V
V_{IK}	Enable input clamp voltage	V_{CC} = 4.5 V, I_I = -5 mA				-1[2]	V
V_{TH+}	Positive-going differential input threshold voltage[1]	$|V_{Rin} - V_{\overline{Rin}}|$	n = 1 or 2			100	mV
V_{TH-}	Negative-going differential input threshold voltage[1]	$|V_{Rin} - V_{\overline{Rin}}|$	n = 1 or 2			-100[2]	mV
V_{HYST}	Differential input threshold voltage hysteresis	$(V_{TH+}$- $V_{TH-})$			50		mV
I_{OZL}	Off-state output low current (high Z)	V_{CC} = 5.5 V, V_O = 0.4 V				-20[2]	µA
I_{OZH}	Off-state output high current (high Z)	V_{CC} = 5.5 V, V_O = 2.4 V				20	µA
I_{OS}	Short circuit output current[3]	V_{CC} = 5.5 V				-100[2]	mA
I_{IL}	Enable input low current	V_{CC} = 5.5 V, V_{IN} = 0.4 V				-400[2]	µA
I_{IH}	Enable input high current	V_{CC} = 5.5 V, V_{IN} = 2.7 V				20	µA
I_{IH}	Enable input reverse current	V_{CC} = 5.5 V, V_{IN} = 5.5 V				100	µA
II_L	Differential input low current	V_{CC} = 5.5V, V_{IN} = -1.2 V				-2[2]	mA
I_{IH}	Differential input high current	V_{CC} = 5.5V, V_{IN} = 7.2 V				1	mA
R_O	Output resistance				20		Ω

(1)   The input levels and difference voltage provide no noise immunity and should be tested only in a static, noise-free environment.
(2)   This parameter is listed using a magnitude and polarity/direction convention, rather than an algebraic convention, to match the original Agere data sheet.
(3)   Test must be performed one lead at a time to prevent damage to the device.

DRIVER SWITCHING CHARACTERISTICS

over operating free-air temperature range unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{P1}	Propagation delay time, input high to output[1]	C_L= 5 pF, See Figure 2 and Figure 6		1.2	2	ns
t_{P2}	Propagation delay time, input low to output[1]			1.2	2	ns
Δt_P	Capacitive delay			0.01	0.03	ns/pF
t_{PHZ}	Propagation delay time, high-level-to-high-impedance output	C_L = 5 pF, See Figure 3 and Figure 6		8	12	ns
t_{PLZ}	Propagation delay time, low-level-to-high-impedance output			7	12	ns
t_{PZH}	Propagation delay time, high-impedance-to-high-level output			4	12	ns
t_{PZL}	Propagation delay time, high-impedance-to-low-level output			5	12	ns
t_{skew1}	Output skew, $|t_{P1} - t_{P2}|$	C_L= 5 pF, See Figure 2 and Figure 6		0.15	0.3	ns
t_{skew2}	Output skew, $|t_{PHH} - t_{PHL}|$, $|t_{PLH} - t_{PLL}|$			0.15	1.1	ns
$t_{skew(pp)}$	Part-to-part skew[2]			0.1	1	ns
Δt_{skew}	Output skew, difference between drivers				0.3	ns
t_{TLH}	Rise time (20%-80%)			0.7	2	ns
t_{THL}	Fall time (80%-20%)			0.7	2	ns

(1)   Parameters t_{P1} and t_{P2} are measured from the 1.5 V point of the input to the crossover point of the outputs (see Figure 2).
(2) $t_{skew(pp)}$ is the magnitude of the difference in propagation delay times between any specified outputs of two devices when both devices operate with the same supply voltage, at the same temperature, and have identical packages and test circuits.

RECEIVER SWITCHING CHARACTERISTICS

over operating free-air temperature range unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH}	Propagation delay time, low-to-high-level output	$C_L = 0$ pF[1], See Figure 4 and Figure 8		2.5	4	ns
t_{PHL}	Propagation delay time, high-to-low-level output			2.5	4	
t_{PLH}	Propagation delay time, low-to-high-level output	$C_L = 15$ pF, See Figure 4 and Figure 8		3	5.5	ns
t_{PHL}	Propagation delay time, high-to-low-level output			3	5.5	
t_{PHZ}	Propagation delay time, high-level-to-high-impedance output	$C_L = 5$ pF, See Figure 5 and Figure 9		6	12	ns
t_{PLZ}	Propagation delay time, low-level-to-high-impedance output			6	12	ns
t_{skew1}	Pulse width distortion, $\lvert t_{PHL} - t_{PLH} \rvert$	Load capacitance (C_L) = 10 pF, See Figure 4 and Figure 8			0.7	ns
		Load capacitance (C_L) = 150 pF, See Figure 4 and Figure 8			4	ns
$\Delta t_{skew1pp}$	Part-to-part output waveform skew[2]	$C_L = 10$ pF, $T_A = 75°C$, See Figure 4 and Figure 8		0.8	1.4	ns
		$C_L = 10$ pF, $T_A = -40°C$ to $85°C$, See Figure 4 and Figure 8			1.5	ns
Δt_{skew}	Same part output waveform skew[2]	$C_L = 10$ pF, See Figure 4 and Figure 8			0.3	ns
t_{PZH}	Propagation delay time, high-impedance-to-high-level output	$C_L = 10$ pF, See Figure 5 and Figure 8		3	12	ns
t_{PZL}	Propagation delay time, high-impedance-to-low-level output			4	12	ns
t_{TLH}	Rise time (20%—80%)	$C_L = 10$ pF, See Figure 5 and Figure 8	1		4	ns
t_{THL}	Fall time (80%—20%)		1		4	ns

(1) The propagation delay values with a 0 pF load are based on design and simulation.
(2) Output waveform skews are when devices operate with the same supply voltage, same temperature, have the same packages and the same test circuits.
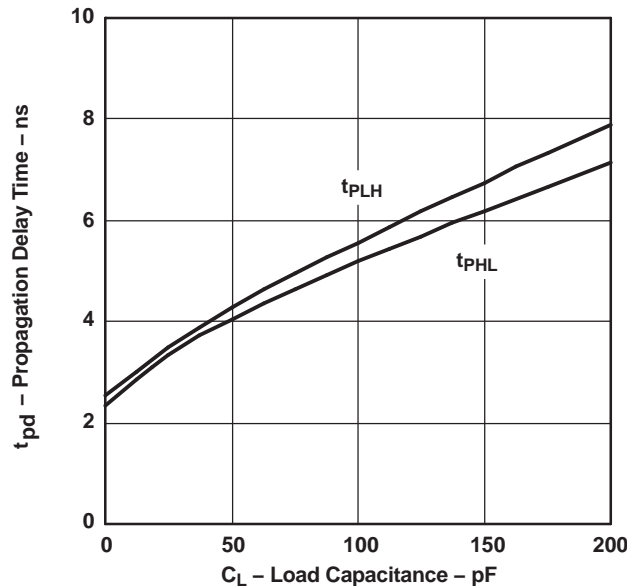
NOTE: This graph is included as an aid to the system designers. Total circuit delay varies with load capacitance. The total delay is the sum of the delay due to external capacitance and the intrinsic delay of the device. Intrinsic delay is listed in the table above as the 0 pF load condition. The incremental increase in delay between the 0 pF load condition and the actual total load capacitance represents the extrinsic, or external delay contributed by the load.

Figure 1. Typical Propagation Delay vs Load Capacitance at 25°C

Product Folder Link(s): *TB5T1*
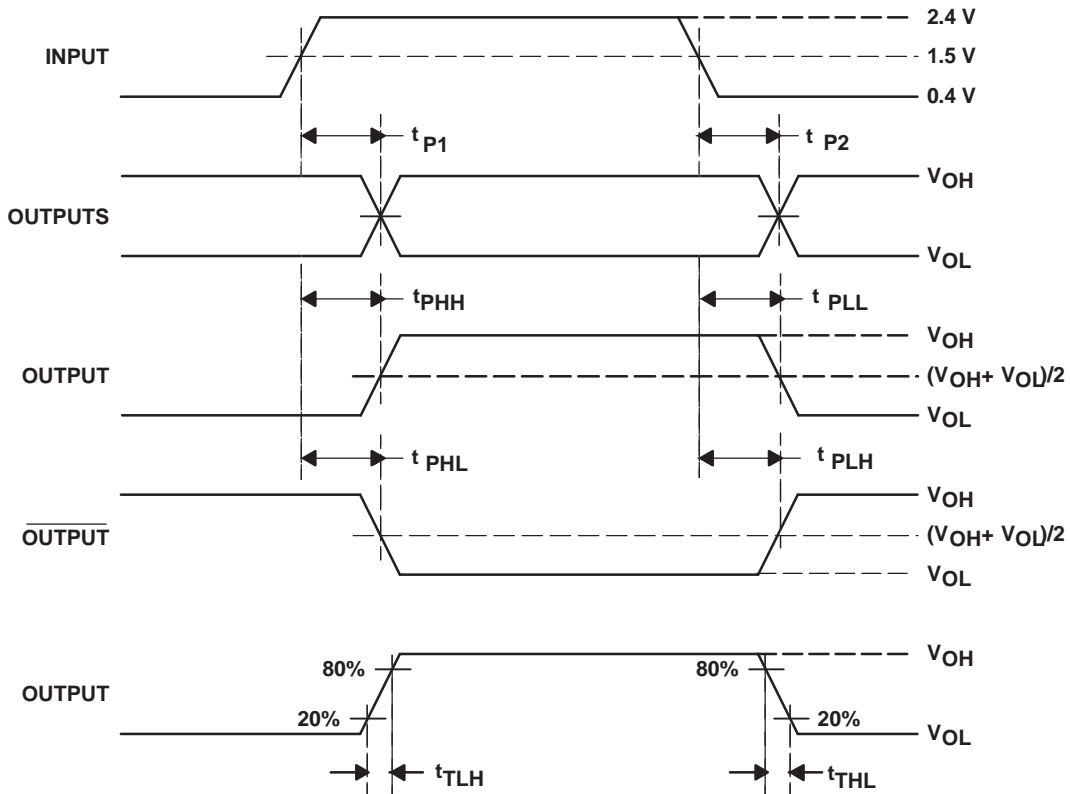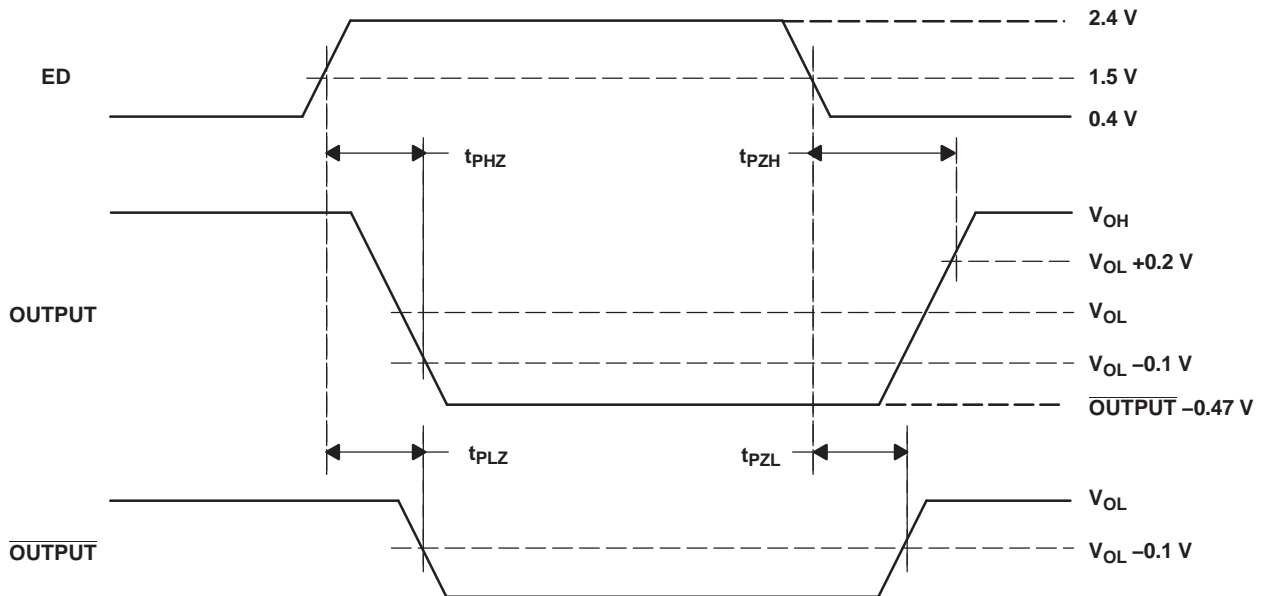
PARAMETER MEASUREMENT INFORMATION

Figure 2. Driver Propagation Delay Times

A.   NOTE: In the third state, OUTPUT is 0.47 V (minimum) more negative than $\overline{\text{OUTPUT}}$.

Figure 3. Driver Enable and Disable Delay Times for a High Input
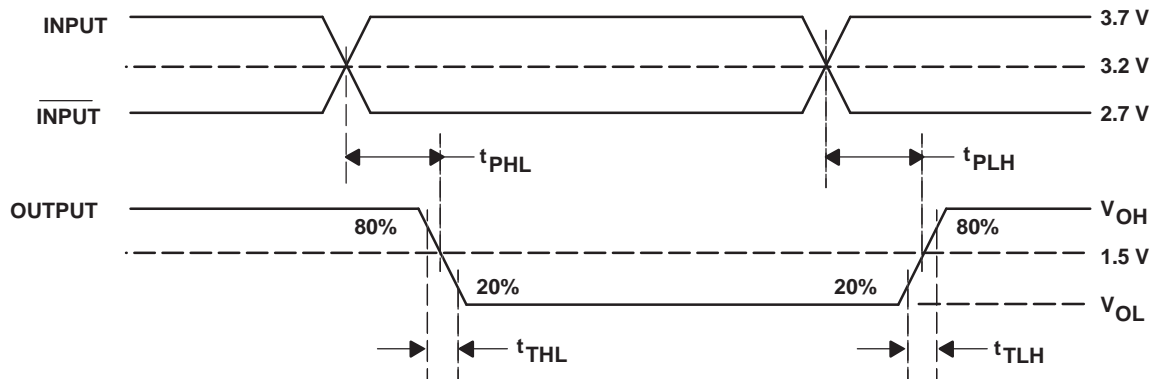
PARAMETER MEASUREMENT INFORMATION (continued)
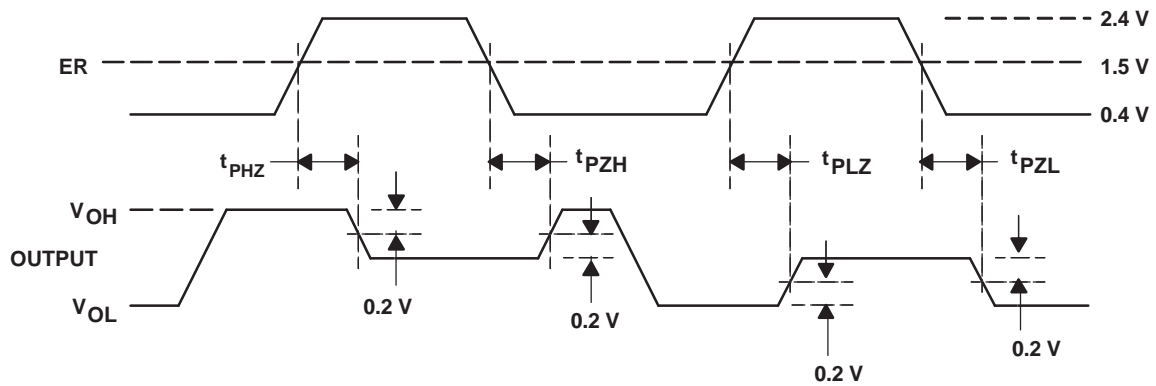

Figure 4. Receiver Propagation Delay Times

Figure 5. Receiver Enable and Disable Timing

Parametric values specified under the Electrical Characteristics and Timing Characteristics sections for the data transmission driver devices are measured with the following output load circuits.
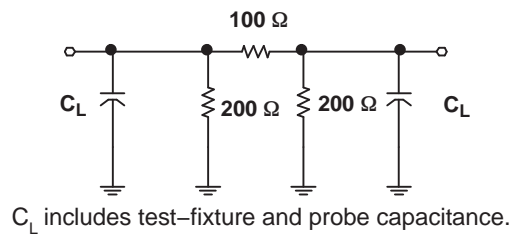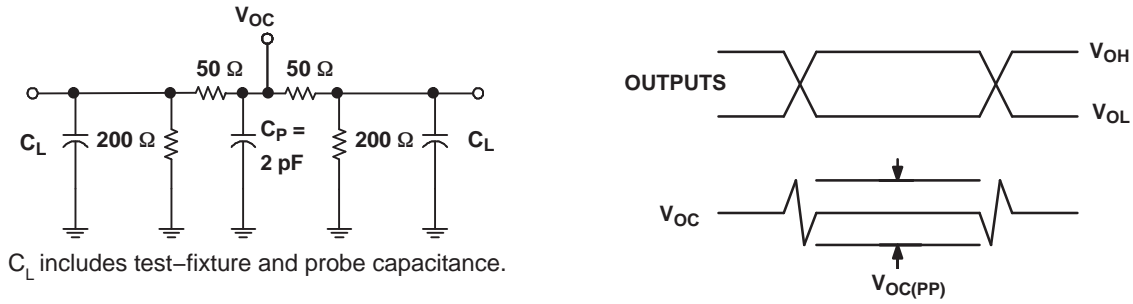

C_L includes test−fixture and probe capacitance.

Figure 6. Driver Test Circuit

Product Folder Link(s): *TB5T1*

PARAMETER MEASUREMENT INFORMATION (continued)

C_L includes test−fixture and probe capacitance.

A.   NOTE: All input pulses are supplied by a generator having the following characteristics: t_r or t_f = 1 ns, pulse repetition rate (PRR) = 0.25 Mbps, pulse width = 500 ± 10 ns. C_P includes the instrumentation and fixture capacitance within 0,06 m of the D.U.T. The measurement of $V_{OS(PP)}$ is made on test equipment with a -3 dB bandwidth of at least 1 GHz.

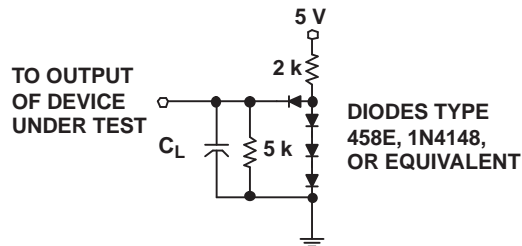**Figure 7. Test Circuit and Definitions for the Driver Common-Mode Output Voltage**

Figure 8. Receiver Propagation Delay Time and Enable Time (t_{PZH}, t_{PZL}) Test Circuit
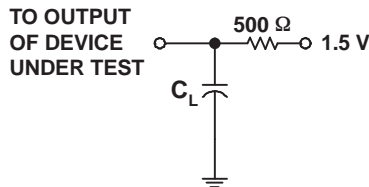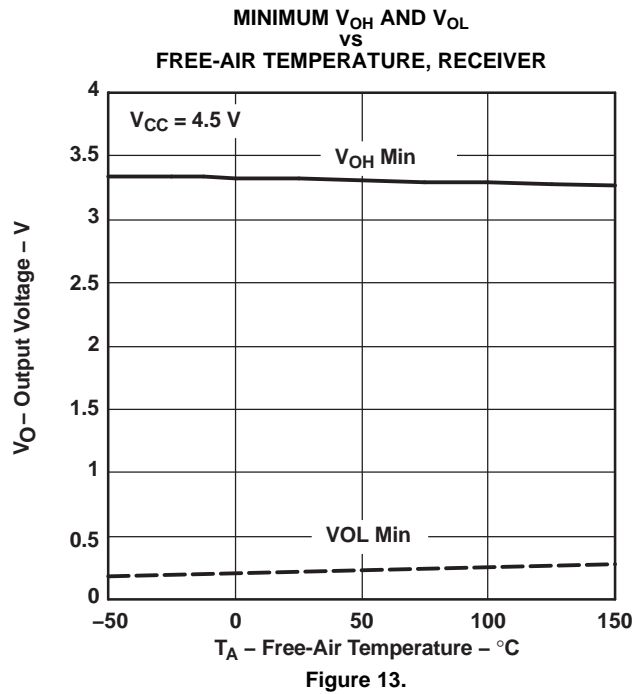

Figure 9. Receiver Disable Time (t_{PHZ}, t_{PLZ}) Test Circuit

Product Folder Link(s): *TB5T1*

TYPICAL CHARACTERISTICS

OUTPUT-VOLTAGE
vs
OUTPUT CURRENT, DRIVER

Figure 10.

V_{OL} AND V_{OH} EXTREMES
vs
FREE-AIR TEMPERATURE, DRIVER

Figure 11.

DIFFERENTIAL OUTPUT VOLTAGE
vs
FREE-AIR TEMPERATURE, DRIVER

Figure 12.

MINIMUM V_{OH} AND V_{OL}
vs
FREE-AIR TEMPERATURE, RECEIVER

Figure 13.

Product Folder Link(s): *TB5T1*

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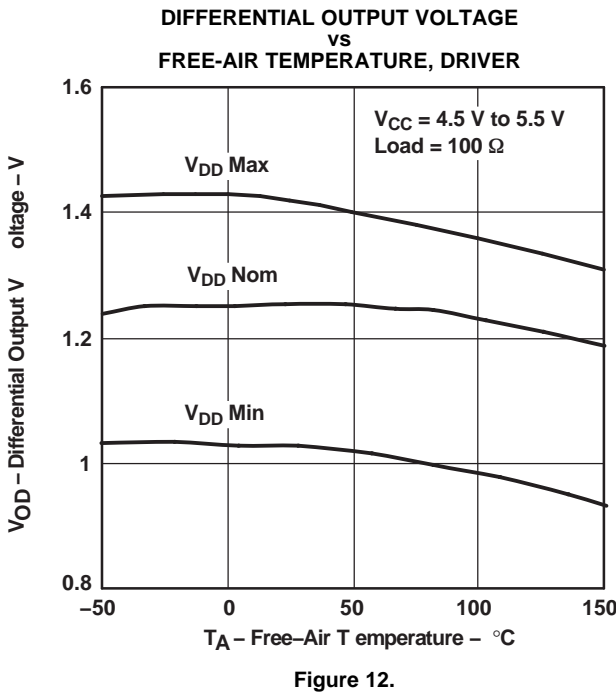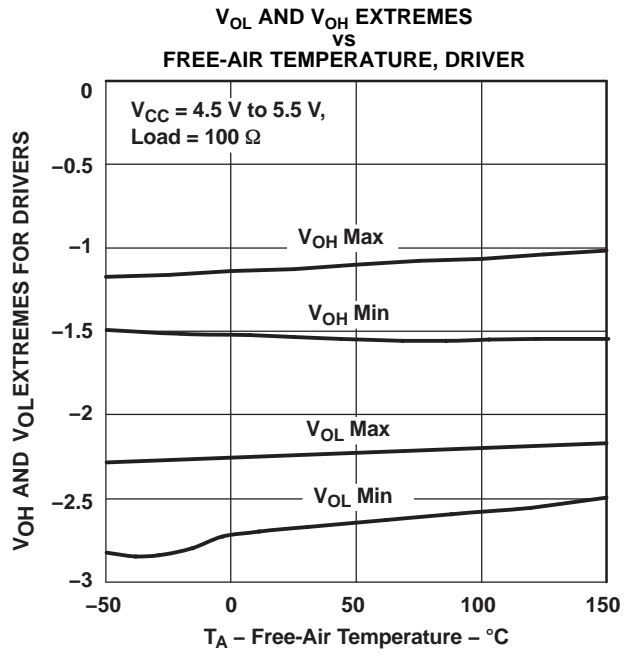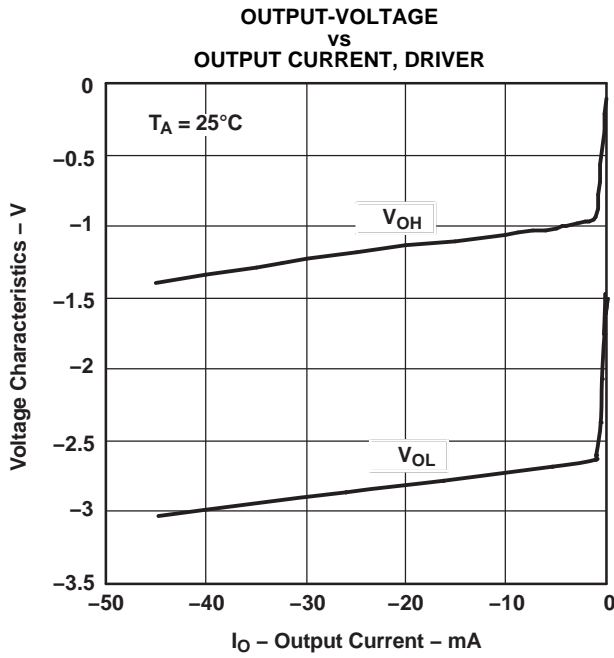
TYPICAL CHARACTERISTICS (continued)

PROPAGATION DELAY TIME t$_{P1}$ or t$_{P2}$
vs
FREE-AIR TEMPERATURE, DRIVER

Figure 14.

LOW-TO-HIGH PROPAGATION DELAY
vs
FREE-AIR TEMPERATURE, RECEIVER

Figure 15.

HIGH-TO-LOW PROPAGATION DELAY
vs
FREE-AIR TEMPERATURE, RECEIVER

Figure 16.

Product Folder Link(s): *TB5T1*
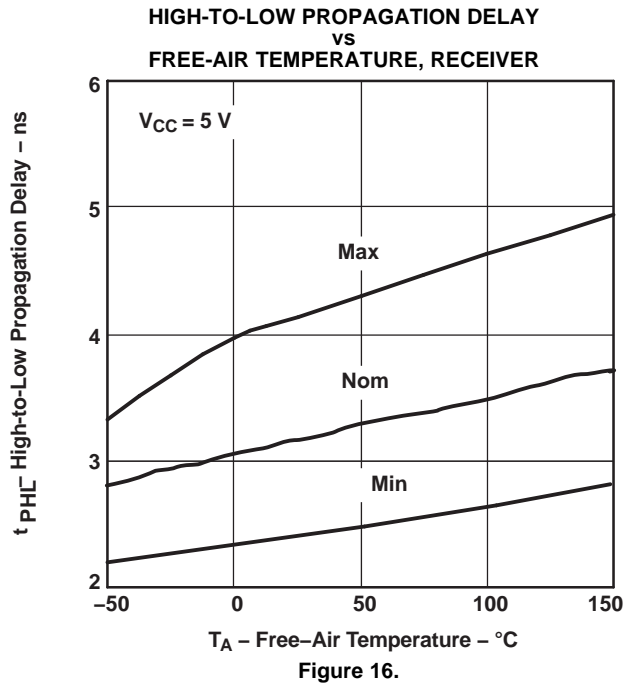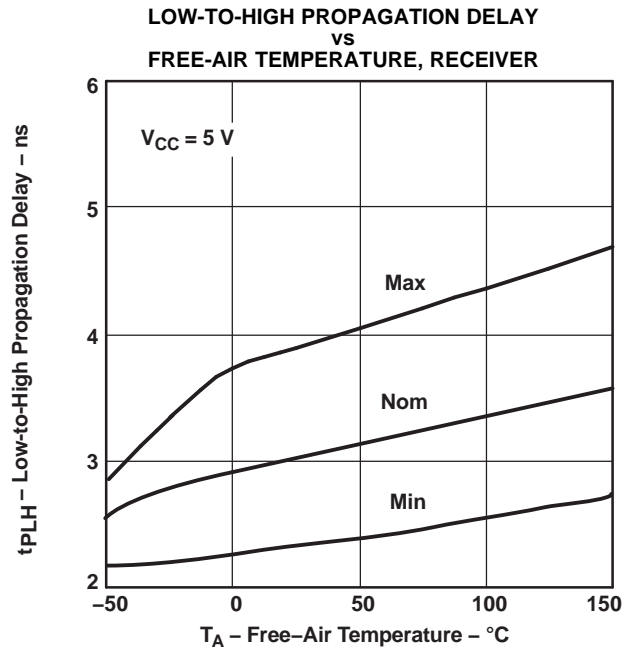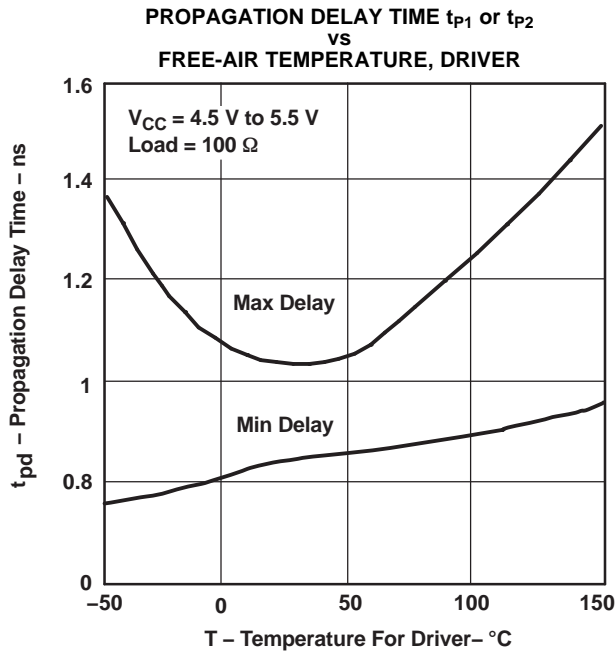
APPLICATION INFORMATION

Power Dissipation

The power dissipation rating, often listed as the package dissipation rating, is a function of the ambient temperature, T_A, and the airflow around the device. This rating correlates with the device's maximum junction temperature, sometimes listed in the absolute maximum ratings tables. The maximum junction temperature accounts for the processes and materials used to fabricate and package the device, in addition to the desired life expectancy.

There are two common approaches to estimating the internal die junction temperature, T_J. In both of these methods, the device internal power dissipation P_D needs to be calculated This is done by totaling the supply power(s) to arrive at the system power dissipation:

$$\sum \left(V_{Sn} \times I_{Sn} \right)$$

and then subtracting the total power dissipation of the external load(s):

$$\sum \left(V_{Ln} \times I_{Ln} \right)$$

The first T_J calculation uses the power dissipation and ambient temperature, along with one parameter: θ_{JA}, the junction-to-ambient thermal resistance, in degrees Celsius per watt.

The product of P_D and θ_{JA} is the junction temperature rise above the ambient temperature. Therefore:

$$T_J = T_A + \left(P_D \times \theta_{JA} \right)$$

Note that θ_{JA} is highly dependent on the PCB on which the device is mounted and on the airflow over the device and PCB. JEDEC/EIA has defined standardized test conditions for measuring θ_{JA}. Two commonly used conditions are the low-K and the high-K boards, covered by EIA/JESD51-3 and EIA/JESD51-7 respectively. Figure 17 shows the low-K and high-K values of θ_{JA} versus air flow for this device and its package options.
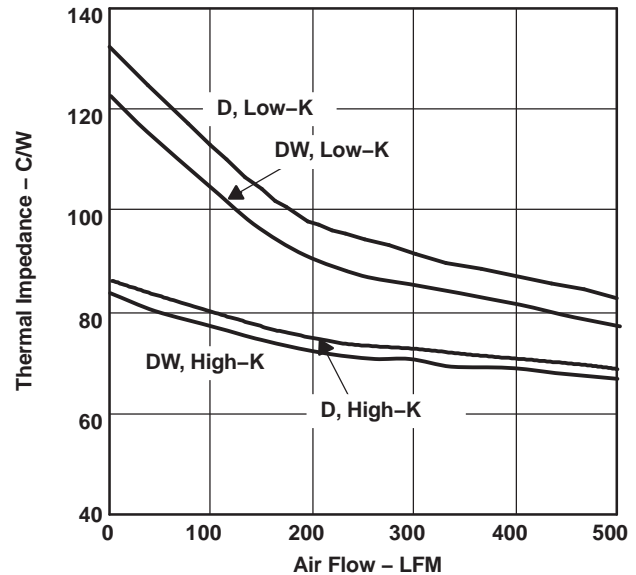

Figure 17. Thermal Impedance vs Air Flow

The standardized θ_{JA} values may not accurately represent the conditions under which the device is used. This can be due to adjacent devices acting as heat sources or heat sinks, to nonuniform airflow, or to the system PCB having significantly different thermal characteristics than the standardized test PCBs. The second method of system thermal analysis is more accurate. This calculation uses the power dissipation and ambient temperature, along with two device and two system-level parameters:

- θ_{JC}, the junction-to-case thermal resistance, in degrees Celsius per watt
- θ_{JB}, the junction-to-board thermal resistance, in degrees Celsius per watt
- θ_{CA}, the case-to-ambient thermal resistance, in degrees Celsius per watt
- θ_{BA}, the board-to-ambient thermal resistance, in degrees Celsius per watt.

In this analysis, there are two parallel paths, one through the case (package) to the ambient, and another through the device to the PCB to the ambient. The system-level junction-to-ambient thermal impedance, $\theta_{JA(S)}$, is the equivalent parallel impedance of the two parallel paths:

$$T_J = T_A + \left(P_D \times \theta_{JA(S)} \right)$$

where

$$\theta_{JA(S)} = \frac{\left[\left(\theta_{JC} + \theta_{CA} \right) \times \left(\theta_{JB} + \theta_{BA} \right) \right]}{\left(\theta_{JC} + \theta_{CA} + \theta_{JB} + \theta_{BA} \right)}$$

Load Circuits

The test load circuits shown in Figure 6 and Figure 7 are based on a recommended *pi* type of load circuit shown in Figure 18. The 100-Ω differential load resistor R_T at the receiver provide proper termination for the interconnecting transmission line, assuming it has a 100-Ω characteristic impedance. The two resistors R_S to ground at the driver end of the transmission line link provide dc current paths for the emitter follower output transistors. The two resistors to ground normally should not be placed at the receiver end, as they shunt the termination resistor, potentially creating an impedance mismatch with undesirable reflections.
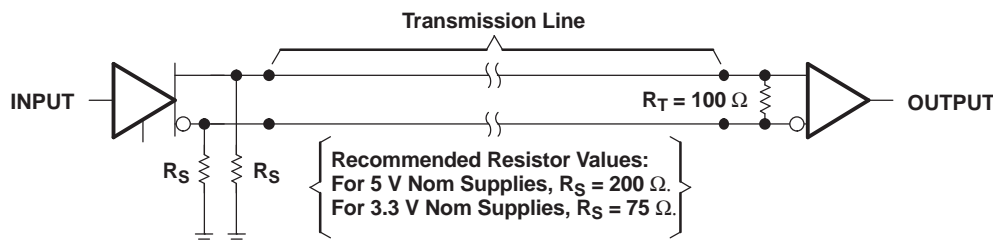

Figure 18. A Recommended pi Load Circuit

Another common load circuit, a Y load, is shown in Figure 19. The receiver-end line termination of R_T is provided by the series combination of the two $R_{T/2}$ resistors, while the dc current path to ground is provided by the single resistor R_S. Recommended values, as a function of the nominal supply voltage range, are indicated in the figure.
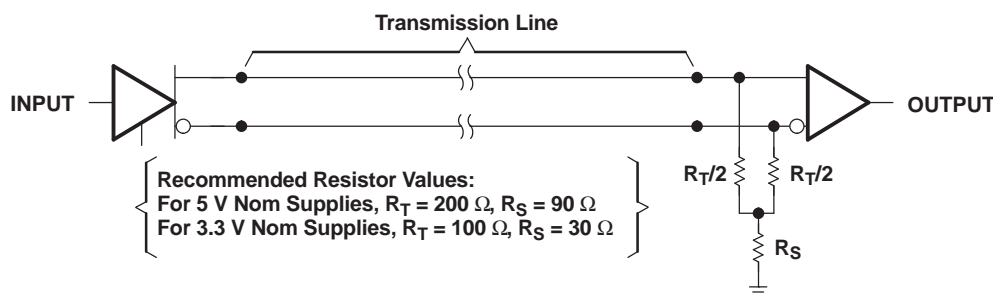
Figure 19. A Recommended Y Load Circuit

An additional load circuit, similar to one commonly used with ECL and PECL, is shown in Figure 20.
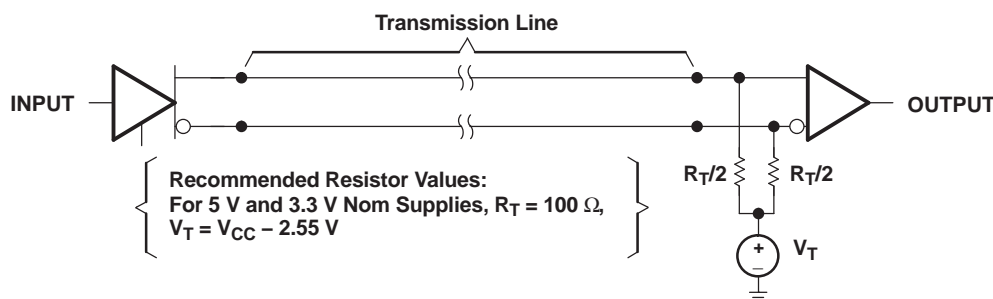
Figure 20. A Recommended PECL-Style Load Circuit

An important feature of all of these recommended load circuits is that they ensure that both of the emitter follower output transistors remain active (conducting current) at all times. When deviating from these recommended values, it is important to make sure that the low-side output transistor does not turn off. Failure to do so increases the t_{skew2} and $V_{OC(PP)}$ values, increasing the potential for electromagnetic radiation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)
TB5T1D	ACTIVE	SOIC	D	16	40	Pb-Free (RoHS)	CU NIPDAU	Level-2-250C-1YEAR/ Level-1-220C-UNLIM
TB5T1DE4	ACTIVE	SOIC	D	16	40	Pb-Free (RoHS)	CU NIPDAU	Level-2-250C-1YEAR/ Level-1-220C-UNLIM
TB5T1DR	ACTIVE	SOIC	D	16	2500	Pb-Free (RoHS)	CU NIPDAU	Level-2-250C-1YEAR/ Level-1-220C-UNLIM
TB5T1DRE4	ACTIVE	SOIC	D	16	2500	Pb-Free (RoHS)	CU NIPDAU	Level-2-250C-1YEAR/ Level-1-220C-UNLIM
TB5T1DW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TB5T1DWE4	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TB5T1DWR	ACTIVE	SOIC	DW	16	2000	Pb-Free (RoHS)	CU NIPDAU	Level-2-250C-1YEAR/ Level-1-220C-UNLIM
TB5T1DWRE4	ACTIVE	SOIC	DW	16	2000	Pb-Free (RoHS)	CU NIPDAU	Level-2-250C-1YEAR/ Level-1-220C-UNLIM

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION

REEL DIMENSIONS

Reel Diameter

Reel Width (W1)

TAPE DIMENSIONS

K0 P1 B0 W

Cavity A0

A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

Sprocket Holes

Q1 Q2 Q3 Q4 Q1 Q2 Q3 Q4

User Direction of Feed

Pocket Quadrants

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TB5T1DR	SOIC	D	16	2500	330.0	16.4	6.5	10.3	2.1	8.0	16.0	Q1
TB5T1DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1

PACKAGE MATERIALS INFORMATION

TAPE AND REEL BOX DIMENSIONS

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TB5T1DR	SOIC	D	16	2500	346.0	346.0	33.0
TB5T1DWR	SOIC	DW	16	2000	346.0	346.0	33.0

D (R−PDSO−G16)   PLASTIC SMALL OUTLINE

$\frac{0.394\ (10,00)}{0.386\ (9,80)}$

⚠C

16   9

$\frac{0.244\ (6,20)}{0.228\ (5,80)}$

$\frac{0.157\ (4,00)}{0.150\ (3,80)}$ ⚠D

Pin 1
Index Area

1   8

$\frac{0.020\ (0,51)}{0.012\ (0,31)}$

0.050 (1,27)

⊕ | 0.010 (0,25) Ⓜ

$\frac{0.010\ (0,25)}{0.004\ (0,10)}$

0.069 (1,75) Max

$\frac{0.010\ (0,25)}{0.005\ (0,13)}$

△ | 0.004 (0,10)

Gauge Plane

Seating Plane

0.010 (0,25)   0°−8°

$\frac{0.050\ (1,27)}{0.016\ (0,40)}$
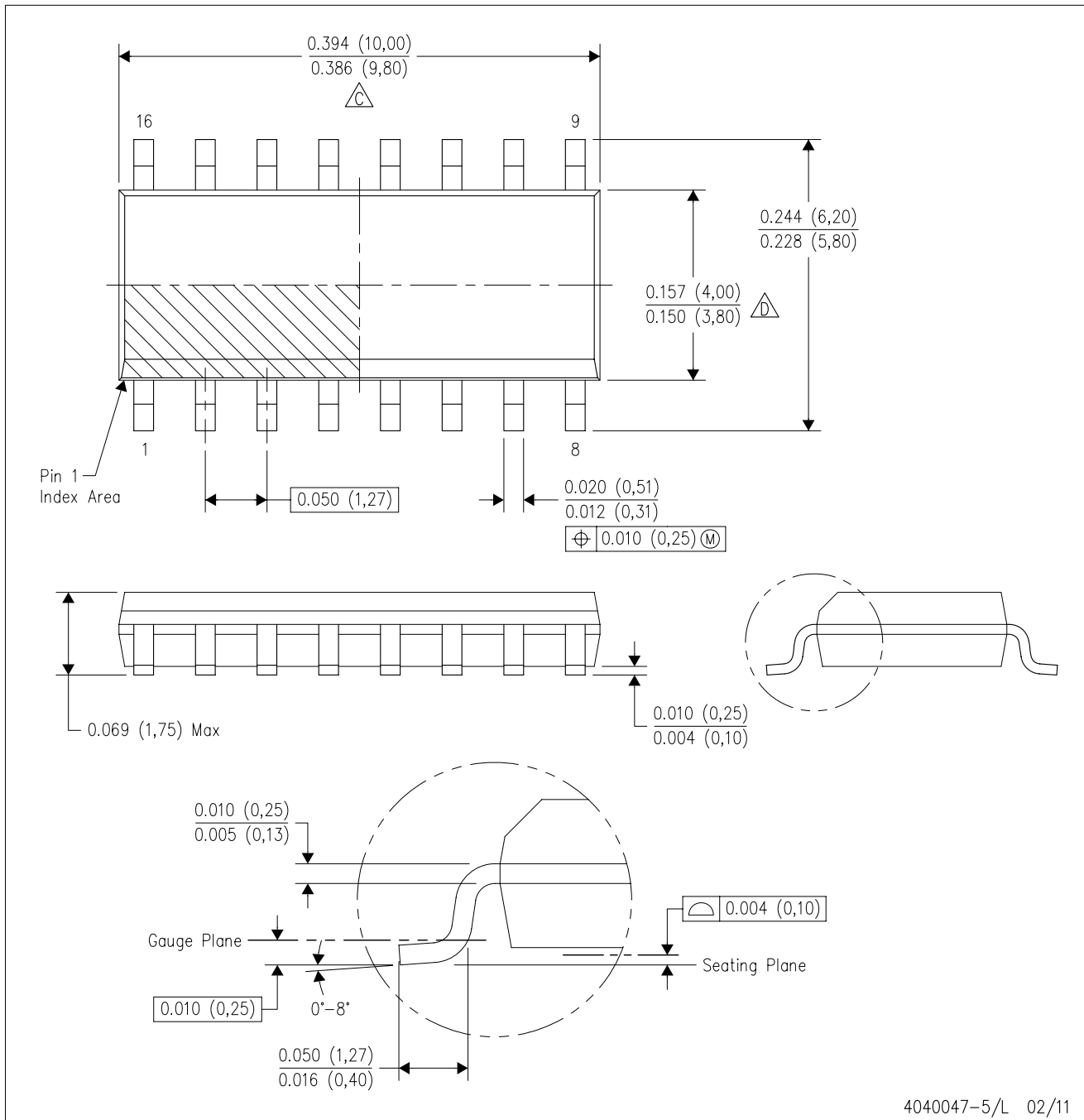
4040047−5/L   02/11

NOTES: A.   All linear dimensions are in inches (millimeters).
         B.   This drawing is subject to change without notice.
        ⚠C   Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall
              not exceed 0.006 (0,15) each side.
        ⚠D   Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
         E.   Reference JEDEC MS−012 variation AC.

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D (R−PDSO−G16)   PLASTIC SMALL OUTLINE

Example Board Layout
(Note C)

Stencil Openings
(Note D)

14x1,27

16x0,55

14x1,27

16x1,95

4,80

4,80

Example
Non Soldermask Defined Pad

Example
Pad Geometry
(See Note C)

Example
Solder Mask Opening
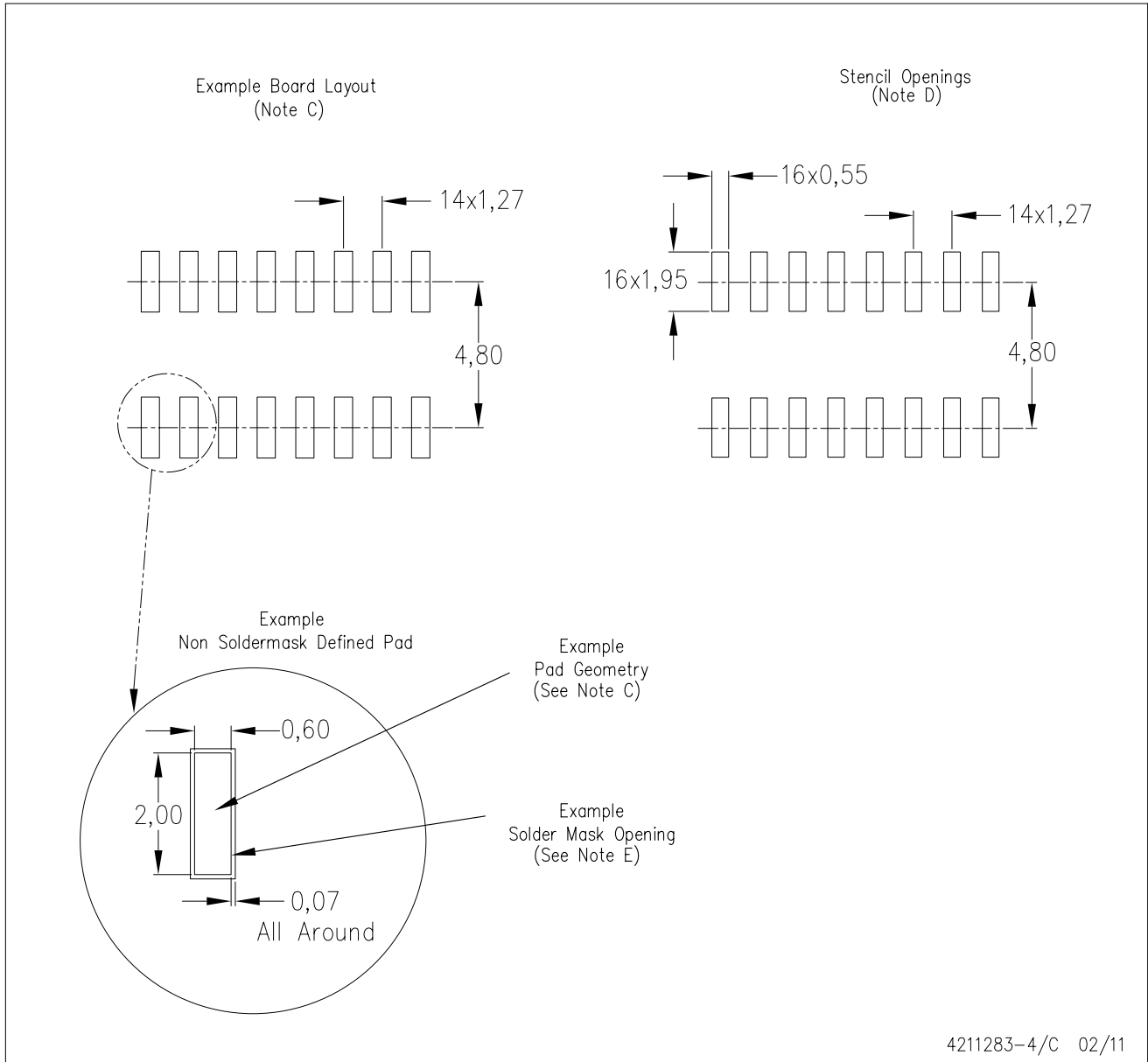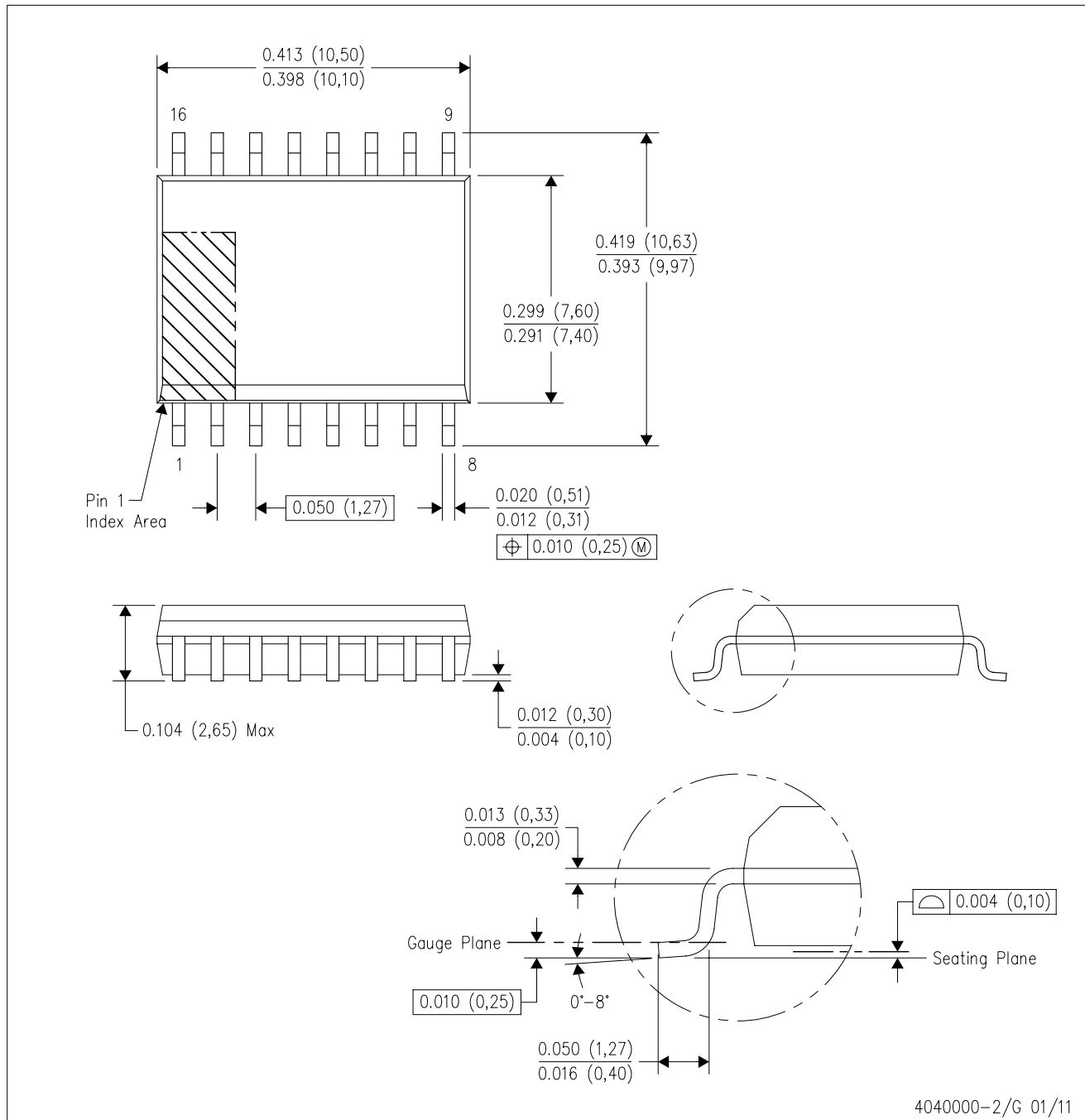(See Note E)

0,60

2,00

0,07
All Around

4211283−4/C 02/11

NOTES: A.   All linear dimensions are in millimeters.
          B.   This drawing is subject to change without notice.
          C.   Publication IPC−7351 is recommended for alternate designs.
          D.   Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should
               contact their board assembly site for stencil design recommendations. Refer to IPC−7525 for other stencil recommendations.
          E.   Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.
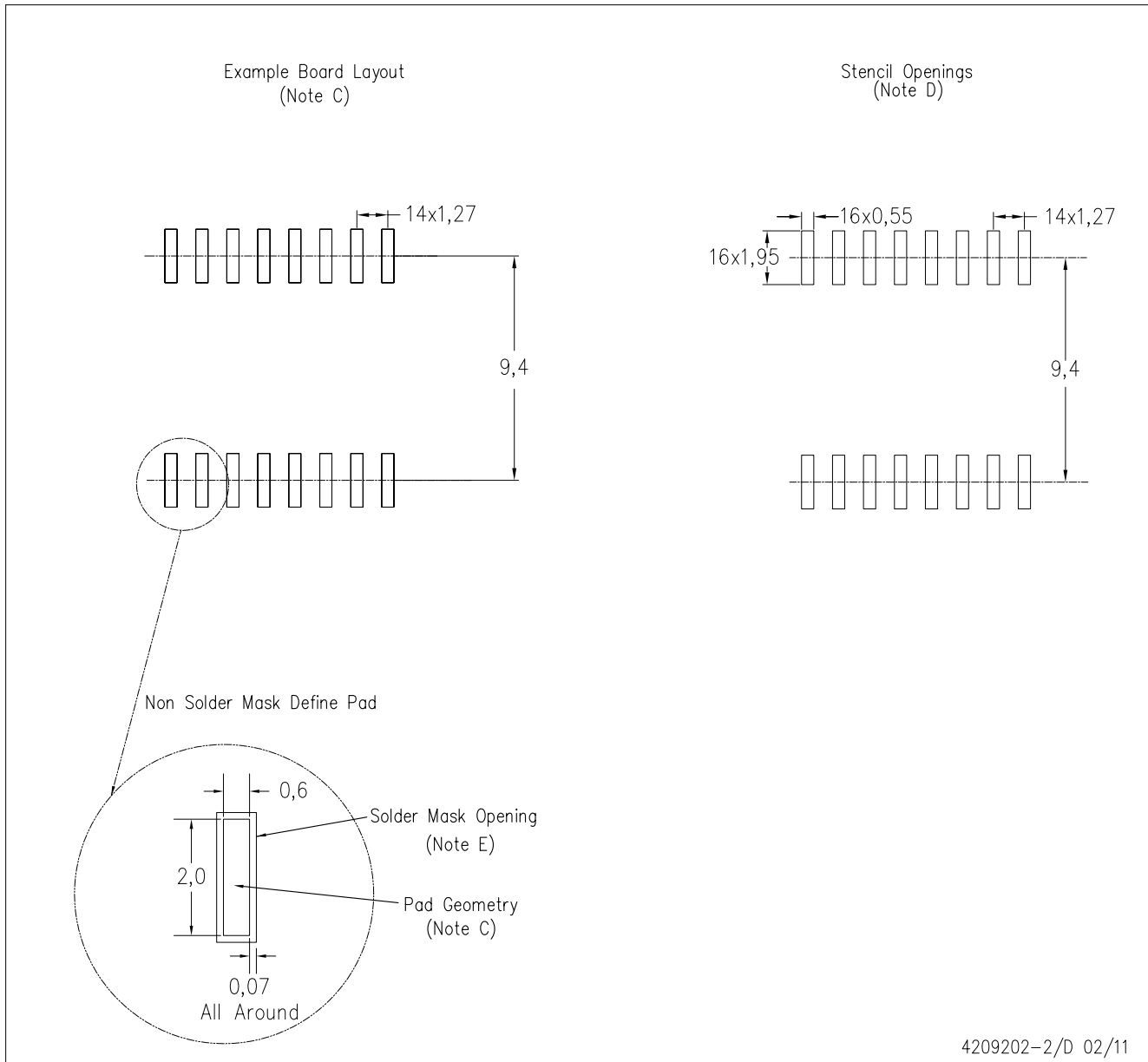
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DW (R−PDSO−G16)   PLASTIC SMALL OUTLINE

0.413 (10,50)
0.398 (10,10)

16   9

0.419 (10,63)
0.393 (9,97)

0.299 (7,60)
0.291 (7,40)

1   8

Pin 1
Index Area

0.050 (1,27)

0.020 (0,51)
0.012 (0,31)

⊕ | 0.010 (0,25) | Ⓜ

0.104 (2,65) Max

0.012 (0,30)
0.004 (0,10)

0.013 (0,33)
0.008 (0,20)

Gauge Plane

0.004 (0,10)

Seating Plane

0.010 (0,25)   0°−8°

0.050 (1,27)
0.016 (0,40)

4040000−2/G 01/11

NOTES: A.   All linear dimensions are in inches (millimeters). Dimensioning and tolerancing per ASME Y14.5M−1994.
         B.   This drawing is subject to change without notice.
         C.   Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
         D.   Falls within JEDEC MS−013 variation AA.

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DW (R−PDSO−G16)   PLASTIC SMALL OUTLINE

Example Board Layout
(Note C)

Stencil Openings
(Note D)

14x1,27

16x0,55   14x1,27

16x1,95

9,4

9,4

Non Solder Mask Define Pad

0,6

Solder Mask Opening
(Note E)

2,0

Pad Geometry
(Note C)

0,07
All Around

4209202−2/D 02/11

NOTES: A.   All linear dimensions are in millimeters.
          B.   This drawing is subject to change without notice.
          C.   Refer to IPC7351 for alternate board design.
          D.   Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should
                contact their board assembly site for stencil design recommendations. Refer to IPC−7525
          E.   Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

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