

# 26V Step-Up Converters for Two to Six White LEDs

#### DESCRIPTION

The EUP2561 is a constant current step-up converter specially designed for driving white LEDs. The step-up converter topology allows series connection of the white LEDs so the LED currents are identical for uniform brightness. The EUP2561 can drive 6 LEDs in series. The brightness of the LEDs can be adjusted through a voltage level on the CTRL pin or by applying a PWM signal to CRTL pin.

1MHz current-mode, pulse-width modulated (PWM) operation allows for small input and output capacitors and a small inductor while minimizing ripple on the input supply/battery. Programmable soft-start eliminates inrush current during startup.

The EUP2561 is available in a space-saving, 8-pin 3mm x 3mm DFN package.

## **FEATURES**

- 2.6V to 5.5V Input Range
- 26V (max) Output with Overvoltage Protection
- Up to 87% Efficiency
- Flexible Analog or PWM Dimming Control
- Internal High Power 30V MOSFET Switch
- < 1μA shutdown Current</p>
- Fast 1MHz PWM Operation
- Small, Low-Profile Inductors and Capacitors
- 3mm × 3mm DFN-8 Package
- RoHS Compliant and 100% Lead (Pb)-Free

#### **APPLICATIONS**

- Cell Phones and Smart Phones
- PDAs, Palmtops, and Wireless Handhelds
- e-Books and Subnotebooks
- White LED Display Backlighting

## **Typical Application Circuit**

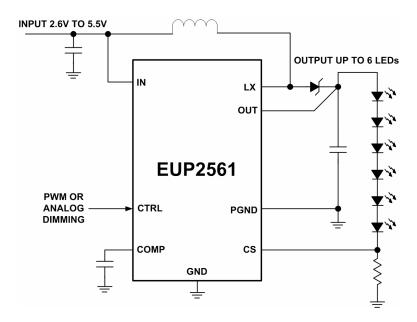


Figure 1. White LED Application



# **Pin Configurations**

Package Type	Pin Configurations			
	(Top View)			
	оит 1	LX		
DFN-8	IN $\boxed{2}$	For the second point of the seco		
	CTRL 3	$\begin{bmatrix} \overline{6} \end{bmatrix}$ GND		
	cs 4]			

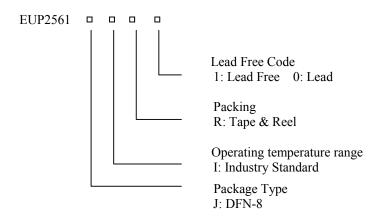
# **Pin Description**

PIN	Pin	DESCRIPTION
OUT	1	Regulated Output Voltage.
IN	2	Input Supply Voltage.
CTRL	3	Brightness Control Input. LED brightness is controlled by the voltage applied to CTRL. Varying the voltage from 0.24V to 1.62V adjusts the brightness from dim to bright, respectively. Any voltage above 1.62V does not increase brightness.
CS	4	Current-Sense Feedback Input. Connect a resistor from CS to GND to set the LED bias current. The voltage at CS regulates to VCTRL / 10 or 0.162V, whichever is lower.
COMP	5	Compensation Input. Connect a 0.1µF capacitor (CCOMP) from COMP to GND. CCOMP stabilizes the converter and controls soft-start. CCOMP discharges to GND when in shutdown.
GND	6	Common Ground. Connect to PGND and the exposed pad directly under the IC.
PGND	7	Power Ground. Connect to GND and the exposed pad directly under the IC.
LX	8	Inductor Connection. This pin is high impedance during shutdown.



# **Ordering Information**

Order Number	Package Type	Marking	Operating Temperature range
EUP2561JIR1	DFN-8	xxxx 2561A	-40 °C to 125°C



# **Block Diagram**

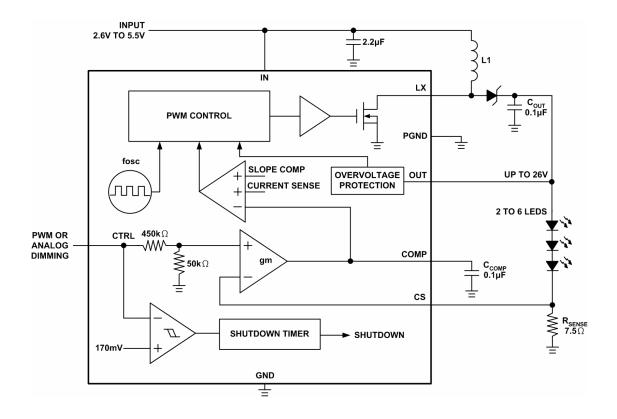


Figure 2.



# **Absolute Maximum Ratings**

	IN to GND	-0.3V to 6V
-	PGND to GND	-0.3V to 0.3V
-	LX,OUT to GND	-0.3V to 30V
-	CTRL to GND $$	or $(V_{IN} + 0.5V)$
-	COMP,CS to GND	to $(V_{IN} + 0.3V)$
-	I <sub>LX</sub>	1A
•	Thermal Resistance	
	DFN-8	50°C /W
-	Junction Temperature	- 150°C
-	Storage Temperature Range	65°C to 150°C
	Lead Temperature (Soldering, 10sec.)	- 260°C

# **Recommended Operating Conditions**

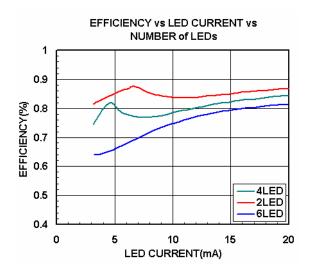
- Supply Voltage,  $V_{IN}$  ------ 2.6V to 5.5V
- **Electrical Characteristics**

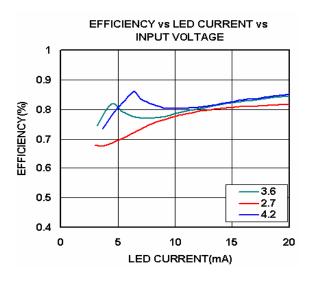
 $(V_{IN}=3V,\,V_{OUT}=20V,\,C_{OUT}=0.1\mu F,\,C_{COMP}=0.15\mu F,\,R_{SENSE}=7.5\Omega,\,V_{CTRL}=1.5V,\,Figure~2,\,T_{A}=-40^{\circ}C~to~+85^{\circ}C,\,unless~otherwise~noted.)$ 

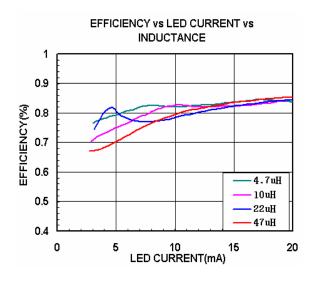
,	Constitution of	]	EUP2561			
Parameter	Conditions	Min	Тур	Max.	Unit	
Supply Voltage		2.6		5.5	V	
UVLO Threshold	V <sub>IN</sub> rising or falling	2.20	2.40	2.60	V	
Quiescent Current	No Switching		0.39	0.60	mA	
OVLO Threshold	V <sub>OUT</sub> rising	26	27.5	29	V	
OUT Input Bias Current	$V_{OUT} = 26V, V_{CTRL} > 0.24V$	12 22		35	4	
OOT input bias current	V <sub>OUT</sub> =V <sub>IN</sub> , CTRL=GND		0.01	1	μA	
Output Voltage Range		$(V_{IN}-V_{D1})$		25.5	V	
ERROR AMPLIFIER						
CTRL to CS Regulation	$V_{CTRL} = 1V$ , $V_{IN} = 2.6V$ to 5.5V	0.095	0.100	0.105	V	
CTRL Input Resistance	$V_{CTRL} \le 1.0V$	200	400	700	kΩ	
CTRL Dual-Mode Threshold	Rising	100	180	260	mV	
CTRL Shutdown Enable Delay			7.5		mS	
CS-to-COMP Transconductance	Vcomp=1.0V	21	52	85	μS	
OSCILLATOR						
Operating Frequency		0.78	1.05	1.3	MHz	
Maximum Duty Cycle	CTRL=IN CS=GND	88	92	96	%	
N-CHANNEL SWITCH						
LX On-Resistance			1.6	2.4	Ω	
LX Current Limit	Duty cycle=90%	550	750	1000	mA	

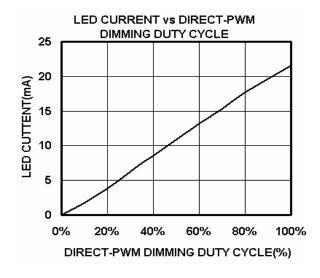


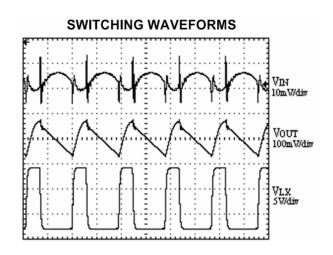
# **Typical Operating Characteristics**

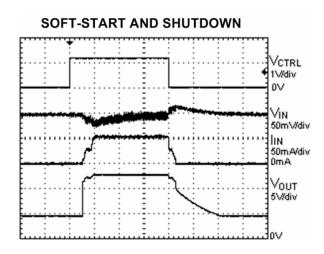






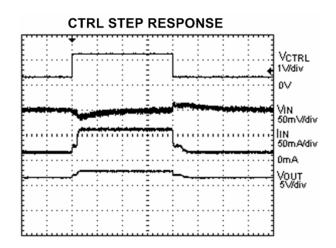


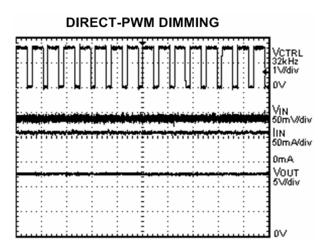




DS2561 Ver1.0 Dec. 2006









# **Application Information**

#### **Soft-Start**

The EUP2561 attain soft-start by charging  $C_{COMP}$  gradually with a current source. When  $V_{COMP}$  rises above 1.25V, the internal MOSFET begins switching at a reduced duty cycle. When  $V_{COMP}$  rises above 2.25V, the duty cycle is at its maximum. See the Typical Operating Characteristics for an example of soft-start operation.

#### Shutdown

The EUP2561 enter shutdown when  $V_{CTRL}$  is less than 8.2ms. In shutdown, 100mV for more than supply current is reduced to 0.3µA by powering down the entire IC except for the CTRL voltage-detection circuitry. C<sub>COMP</sub> is discharged during shutdown, allowing the device to reinitiate soft-start when it is enabled. Although the internal N-channel MOSFET does not switch in shutdown, there is still a DC current path between the input and the LEDs through the inductor and Schottky diode. The minimum forward voltage of the LED array must exceed the maximum input voltage to ensure that the LEDs remain off in shutdown. However, with two or more LEDs, the forward voltage is large enough to keep leakage current low, less than 1μA(typ). Typical shutdown timing characteristics are shown in the Typical Operating Characteristics.

#### **Overvoltage Protection**

Overvoltage lockout (OVLO) occurs when  $V_{OUT}$  is above 27V. The protection circuitry stops the internal MOSFET from switching and causes  $V_{COMP}$  to decay to 0V. The device comes out of OVLO and into softstart when  $V_{OUT}$  falls below 25V.

## **Adjusting LED Current**

Adjusting the EUP2561 output current changes the brightness of the LEDs. An analog input (CTRL) and the sense-resistor value set the output current. Output current is given by:

$$I_{\text{LED}} = \frac{V_{\text{CTRL}}}{10 \times R_{\text{SENSE}}}$$

The  $V_{CTRL}$  voltage range for adjusting output current is 0.24V to 1.62V. To set the maximum current, calculate RSENSE when  $V_{CTRL}$  is at its maximum as follows:

$$R_{\text{SENSE}} = \frac{1.62V}{10 \times I_{\text{LED(MAX)}}}$$

Power dissipation in RSENSE is typically less than 5mW, making a standard chip resistor sufficient.

## **PWM Dimming Control**

CTRL is also used as a digital input allowing LED brightness control with a logic-level PWM signal applied directly to CTRL. The frequency range is from 200Hz to 200kHz, while 0% duty cycle corresponds to zero current and 100% duty cycle corresponds to full current. The error amplifier and compensation capacitor form a lowpass filter so PWM dimming results in DC current to the LEDs without the need for any additional RC filters; see the Typical Operating Characteristics.

#### **Capacitor Selection**

The exact values of input and output capacitors are not critical. The typical value for the input capacitor is  $2.2\mu F$  and the typical value for the output capacitor is  $0.1\mu F$ . Larger value capacitors can be used to reduce input and output ripple, but at the expense of size and higher cost.  $C_{COMP}$  stabilizes the converter and controlls soft-start. Connect a  $0.1\mu F$  capacitor from COMP to GND. For stable operation,  $C_{OUT}$  must not exceed 10 times  $C_{COMP}$ .

#### **Inductor Selection**

Inductor values range from  $10\mu H$  to  $47\mu H$ . A  $22\mu H$  inductor optimizes the efficiency for most applications while maintaining low  $15mV_{P-P}$  input ripple. With input voltages near 5V, a larger value of inductance can be more efficient. To prevent core saturation, ensure that the inductor-saturation current rating exceeds the peak inductor current for the application. Calculate the peak inductor current with the following formula:

$$I_{PEAK} = \frac{V_{OUT(MAX)} \times I_{LED(MAX)}}{0.8 \times V_{IN(MIN)}} + \frac{V_{IN(MIN)} \times 0.8 \mu_{S}}{2 \times I_{s}}$$

## **Schottky Diode Selection**

The EUP2561 high switching frequency demands a high-speed rectification diode (D1) for optimum efficiency. A Schottky diode is recommended due to its fast recovery time and low forward-voltage drop. Ensure that the diode's average and peak current rating exceed the average output current and peak inductor current. In addition, the diode's reverse breakdown voltage must exceed  $V_{\rm OUT}$ . The RMS diode current can be calculated from:

$$I_{DIODE(RMS)} \cong \sqrt{I_{OUT} \times I_{PEAK}}$$



## **PC Board Layout**

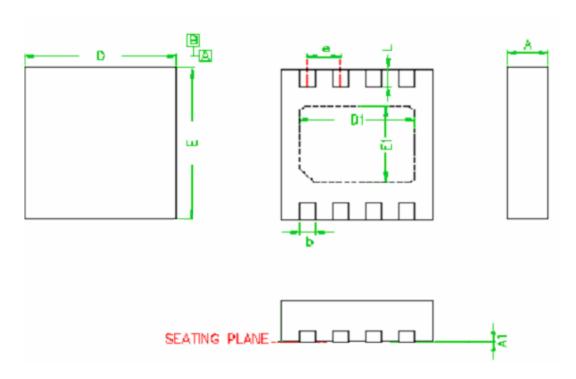
Due to fast switching waveforms and high-current paths, careful PC board layout is required. An evaluation kit is available to speed design.

When laying out a board, minimize trace lengths between the IC and  $R_{\rm SENSE}$ , the inductor, the diode, the input capacitor, and the output capacitor. Keep traces short, direct, and wide. Keep noisy traces, such as the LX node trace, away from CS. The IN bypass capacitor  $(C_{\rm IN})$  should be placed as close to the IC as possible. PGND and GND should be connected directly to the exposed paddle underneath the IC. The ground connections of  $C_{\rm IN}$  and  $C_{\rm OUT}$  should be as close together as possible. The traces from IN to the inductor and from the Schottky diode to the LEDs may be longer.



# **Packaging Information**

DFN-8



SYMBOLS	MILLIMETERS		MILLIMETERS INCHES		
	MIN.	MAX.	MIN.	MAX.	
A	0.70	0.90	0.028	0.035	
A1	0.00	0.05	0.000	0.002	
b	0.20	0.40	0.008	0.016	
D	2.85	3.15	0.112	0.124	
D1	2.30		0.090		
Е	2.85	3.15	0.112	0.124	
E1	1.50		0.059		
e	0.65		0.026		
L	0.25	0.45	0.010	0.018	

